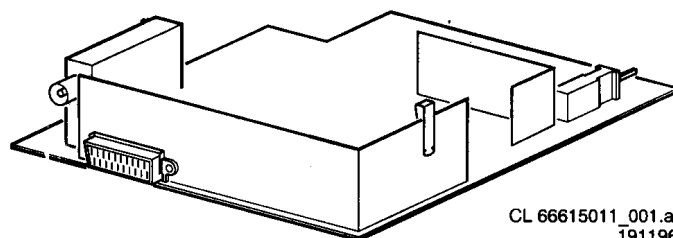


Service  
Service  
Service

**L7.1A**  
AA



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# Service Manual

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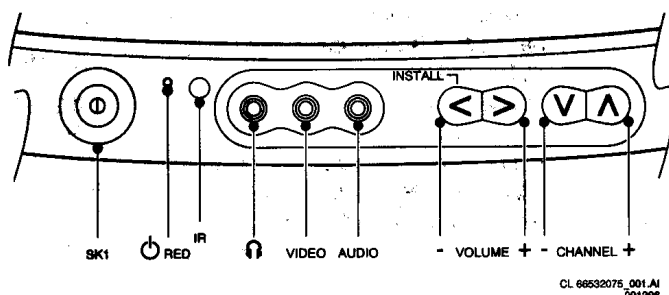
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**PHILIPS**

## 1. Technical specifications

Mains voltage	150 - 276V AC; 50/60 Hz	
	90 - 276V 50/60 Hz (full range)	
Power cons. at 220V~	14" 43W (stand-by ≤ 10W)	
	20" 52W (stand-by ≤ 10W)	
Aerial input impedance TV	75Ω - coax	
Max. aerial input VHF/UHF	100dBμV	
Pull-in range colour sync	± 300Hz	
Pull-in range horizontal sync	± 600Hz	
Pull-in range vertical sync	45 - 64.5 Hz	
Picture tube range	14"	All tubes are universal tube A34EDJ01X024 - LATAM /67 A34JXV70X /59 /50 /71 /97 /75 /73 /93 A34JFQ40X(W) /57 /58 370KSB22 - SYB - A48EEB05X020 - LATAM /75 /73 /58B A48KXR98X /67 A48JRK10X /59 /50 /97 /71 /57 /58 510UFB22 TC69(DPY)
	20"	Universal tubes
	20"	Northern tubes
	14" mono	16Ω 4W front firing loudspeaker
	20" mono	16Ω 3W front firing loudspeaker
TV Systems	/50 /67	PAL B/G
	/75	PAL B/H
	/73 /57	PAL I
	/58 /59	PAL B/GI & SECAM B/G D/K
	/77 /97	NTSC M
	/93	PAL D/I & SECAM D/K
Indications	On Screen Display (OSD) green/red 1 LED (⦿ red high intensity, ⊕ red low intensity, "RC5" and error codes blinking red)	
VCR programs	Any program numbers.	
Tuning and operating system	VST / PLL	
UV1335 /IEC (VST)	Band I	48.25 - 93.25 MHz
	Band III	168.25 - 216.25 MHz
	UHF	471.25 - 863.25 MHz
UV1336 (PLL)	Band I	55.25 - 83.25 MHz
	Band III	175.25 - 211.25 MHz
	UHF	471.25 - 801.25 MHz
Local operating functions	VOLUME + / PROGRAM + /	



## 2. Connection facilities

### Cinch:

- ⊙ CINCH CVBS ⊖ (1V pp +/- 3dB 75 Ω max 2V DC)
- ⊙ CINCH AUDIO ⊖ (500mV RMS < 1K Ω max 2Volt RMS)

### Head phone:

- ⊙ ⊕/⊖ 8 -600Ω/5mW

## Safety instructions for repairs

1. Safety regulations require that **during** a repair:
  - the set should be connected to the mains via an isolating transformer;
  - safety components, indicated by the symbol ▲, should be replaced by components identical to the original ones;
  - when replacing the CRT, safety goggles must be worn.
2. Safety regulations require that **after** a repair the set must be returned in its original condition. In particular attention should be paid to the following points:
  - As a strict precaution, we advise you to resolder the solder joints through which the horizontal deflection current is flowing, in particular:
    - all pins of the line output transformer (LOT);
    - fly-back capacitor(s);
    - S-correction capacitor(s);
    - line output transistor;
    - pins of the connector with wires to the deflection coil;
    - other components through which the deflection current flows.

### Note:

This resoldering is advised to prevent bad connections due to metal fatigue in solder joints and is therefore only necessary for television sets older than 2 years.

- The wire trees and EHT cable should be routed correctly and fixed with the mounted cable clamps.
- The insulation of the mains lead should be checked for external damage.
- The mains lead strain relief should be checked for its function in order to avoid touching the CRT, hot components or heat sinks.
- The electrical DC resistance between the mains plug and the secondary side should be checked (only for sets which have a mains isolated power supply). This check can be done as follows:
  - unplug the mains cord and connect a wire between the two pins of the mains plug;
  - set the mains switch to the on position (keep the mains cord unplugged!);
  - measure the resistance value between the pins of the mains plug and the metal shielding of the tuner or the aerial connection on the set. The reading should be between 4.5 MΩ and 12 MΩ;
  - switch off the TV and remove the wire between the two pins of the mains plug.
- The cabinet should be checked for defects to avoid touching of any inner parts by the customer.

## Maintenance instructions

It is recommended to have a maintenance inspection carried out by a qualified service employee. The interval depends on the usage conditions:

- When the set is used under normal circumstances, for example in a living room, the recommended interval is 3 to 5 years.
- When the set is used in circumstances with higher dust, grease or moisture levels, for example in a kitchen, the recommended interval is 1 year.

The maintenance inspection contains the following actions:

- Execute the above mentioned 'general repair instruction'.
- Clean the power supply and deflection circuitry on the chassis.
- Clean the picture tube panel and the neck of the picture tube.

## Warnings

1. In order to prevent damage to IC's and transistors any flash-over of the EHT should be avoided. To prevent damage to the picture tube the method, indicated in Fig. 3.1, has to be applied to discharge the picture tube. Make use of an EHT probe and a universal meter (position DC-V). Discharge until the reading of the meter is 0V (after approx. 30s).

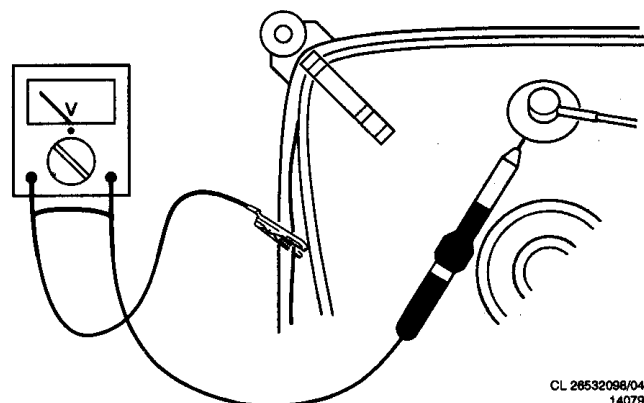


Fig. 3.1

## 2. ESD ▲

All ICs and many other semiconductors are susceptible to electrostatic discharges (ESD). Careless handling during repair can reduce life drastically. When repairing, make sure that you are connected with the same potential as the mass of the set by a wristband with resistance. Keep components and tools also at this same potential.

### Available ESD protection equipment:

anti-static table mat large 1200x650x1.25mm	4822 466 10953
anti-static table mat small 600x650x1.25mm	4822 466 10958
anti-static wristband	4822 395 10223
connection box (3 press stud connections, 1 M ohm)	4822 320 11307
extension cable (2 m, 2 M ohm; to connect wristband to connection box)	4822 320 11305
connecting cable (3 m, 2 M ohm; to connect table mat to connection box)	4822 320 11306
earth cable (1 M ohm; to connect any product to mat or connection box)	4822 320 11308
complete kit ESD3 (combining all 6 prior products - small table mat)	4822 310 10671
wristband tester	4822 344 13999

3. Together with the deflection unit and any multipole unit, the flat square picture tubes used form an integrated unit. The deflection and the multipole units are set optimally at the factory. Adjustment of this unit during repair is therefore not recommended.

4. Proceed with care when testing the EHT section and the picture tube.
5. Never replace any modules or any other parts while the set is switched on.
6. Use plastic instead of metal alignment tools. This will prevent any short circuits and the danger of a circuit becoming unstable.
7. Upon a repair of a transistor or an IC assembly (e.g. a transistor or IC with heatsink and spring) remounting should be carried out in the following order:
  1. Mount transistor or IC on heatsink with spring.
  2. Resolder the joints.

## Notes

1. Do not use heatsinks as earth reference.
2. The direct voltages and oscillograms should be measured with regard to the tuner earth ( $\perp$ ), or hot earth ( $\perp$  with lightning bolt) as this is called.
3. The direct voltages and waveforms are measured in the Service Default Mode (see chapter 8). Use a colour bar pattern of a pattern generator (e.g. PM5518).
4. The DC voltages and oscillograms are where necessary measured with ( $\sqcap$ ) and without ( $\times$ ) aerial signal (settings as in Service Default Mode; see chapter 8). Voltages and oscillograms in the power supply section have been measured for both normal operation ( $\textcircled{I}$ ) and in the stand-by mode ( $\textcircled{C}$ ). As an input signal a colour bar pattern has been used.
5. The picture tube PWB has printed spark gaps. Each spark gap is connected between an electrode of the picture tube and the Aquadag coating.

## 4. Mechanical instructions

For the main carrier two service positions are possible (Fig. 4.1):

- A: For faultfinding on the component side of the main carrier
- B: For (de)soldering activities on the copper side of the main carrier

Position A can be reached by first removing the mains cord from its fixation, then loosen the carrier lips (1) and then pulling the carrier panel (2) for approximately 10 cm.

Position B can be reached from position A after disconnecting the degaussing cable. Put the carrier on the line transformer side and if wanted use a screwdriver for an extra stable service position (see figure below).

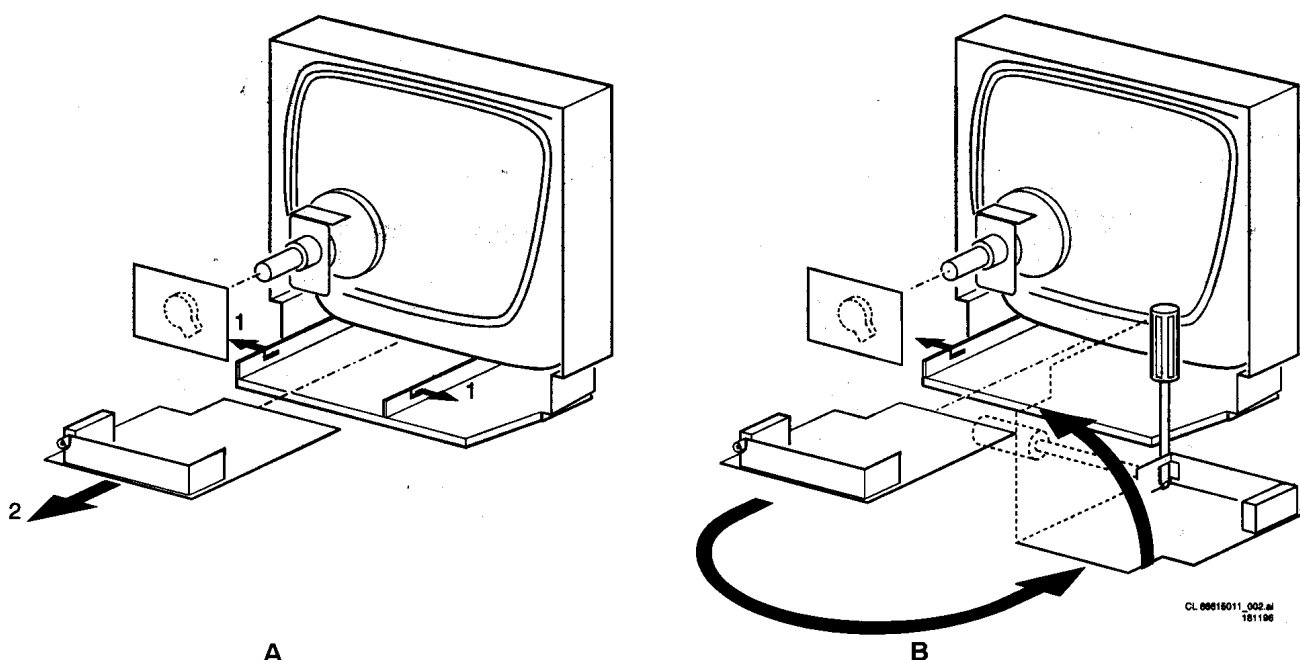
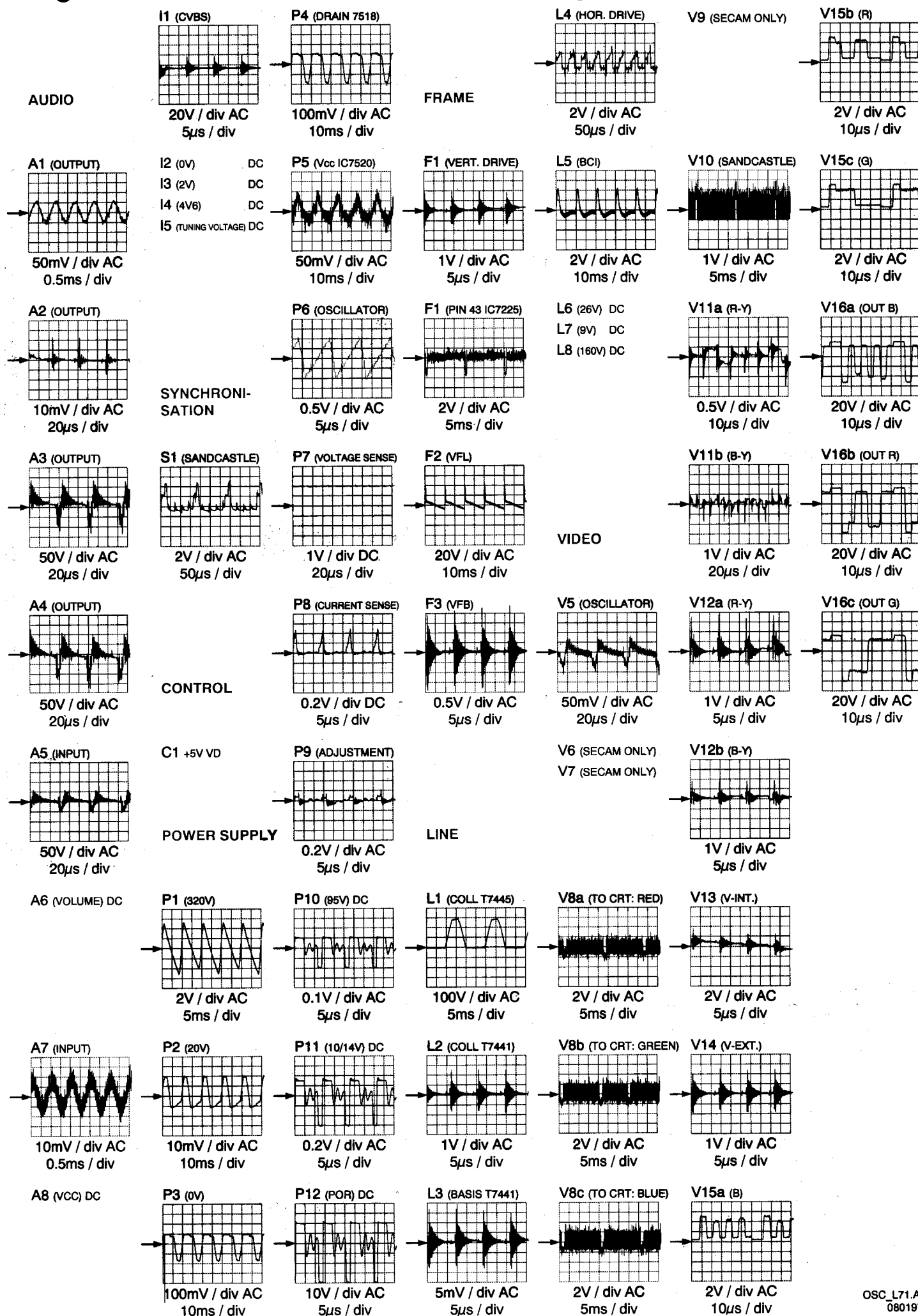


Fig. 4.1

# 5. Overview oscillograms / Übersicht Oszillogramme / Vue d'ensemble des oscillogrammes

Chassis L7.1A

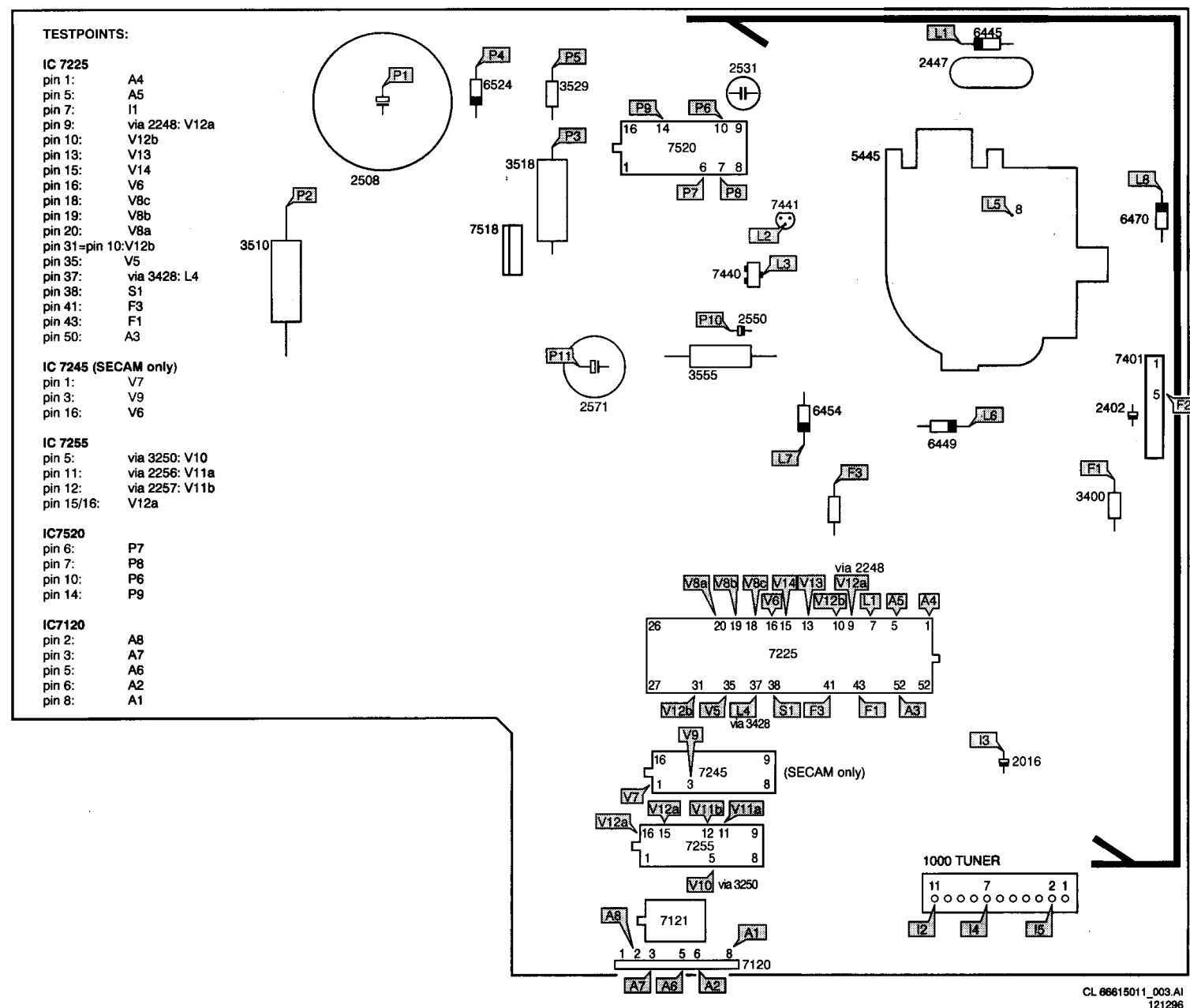
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OSC\_L71.A1  
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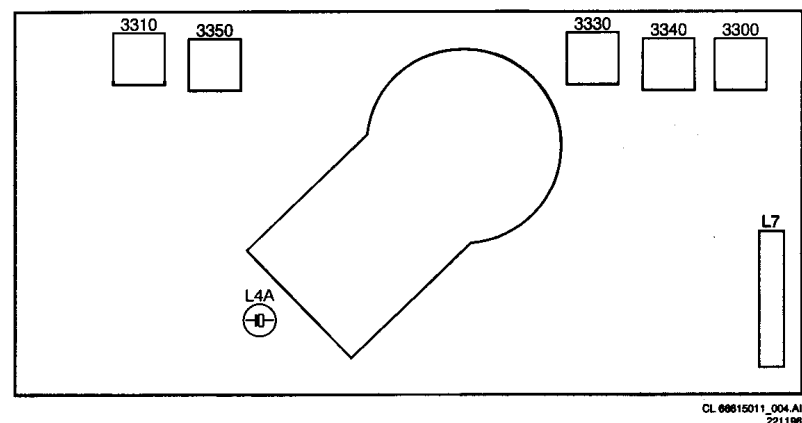
## Survey of testpoints / Übersicht über die Teststellen / Presentation des points à tester

### MAIN CARRIER (Component side)



**Fig. 5.1**

## CRT PANEL

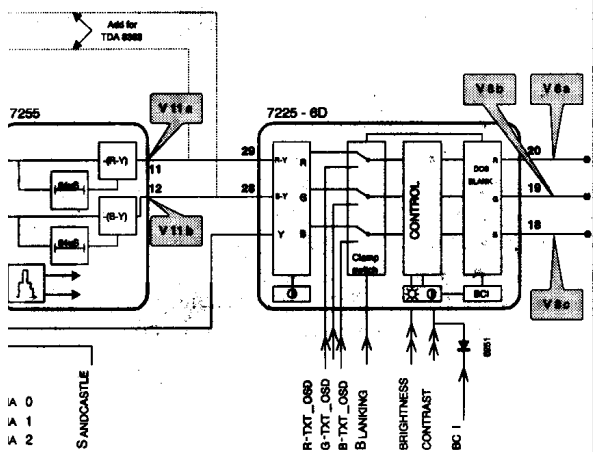


**Fig. 5.2**

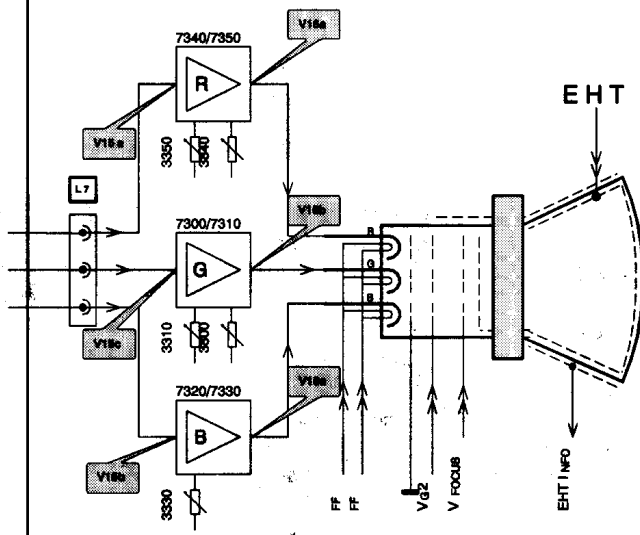


# Diagramme synoptique

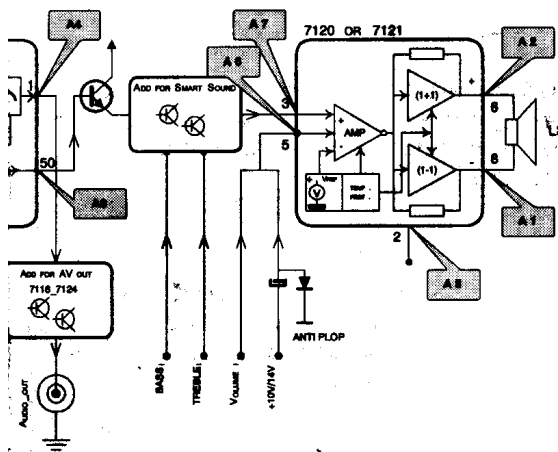
## VIDEO & CHROMA PROCESSING A7



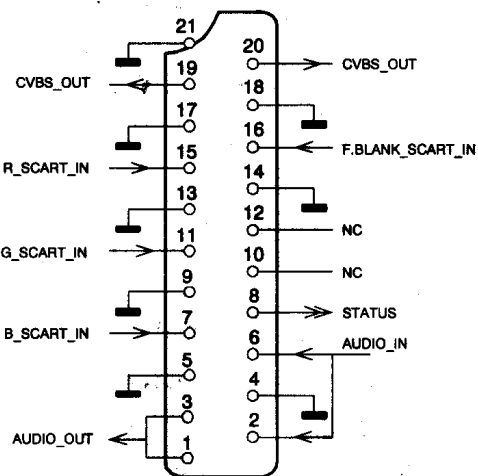
## B1 CRT MODULE



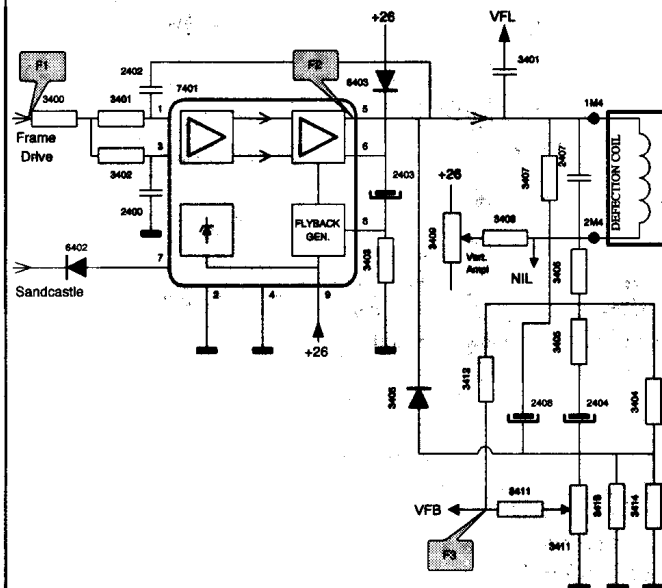
## SOUND AMPLIFIER A8



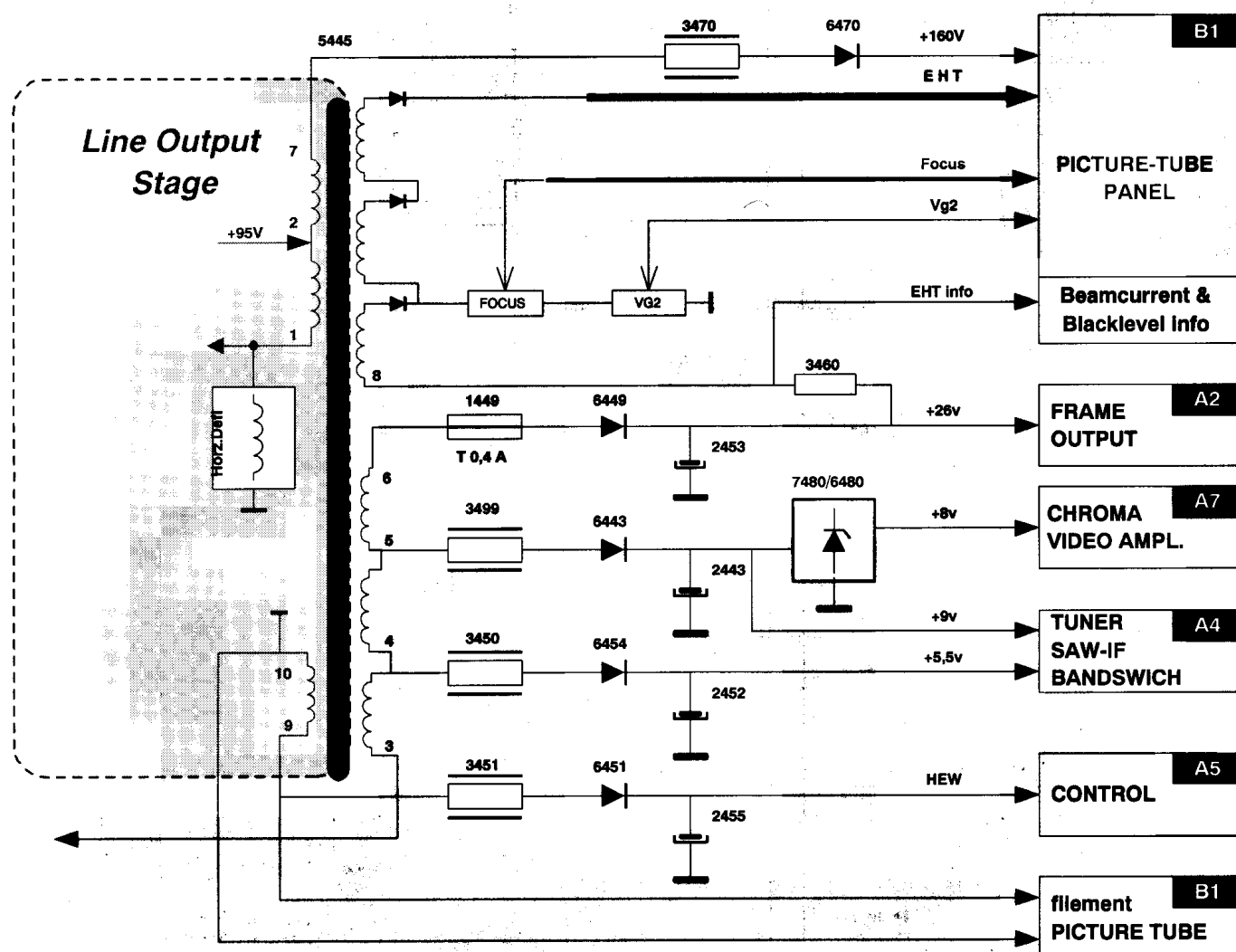
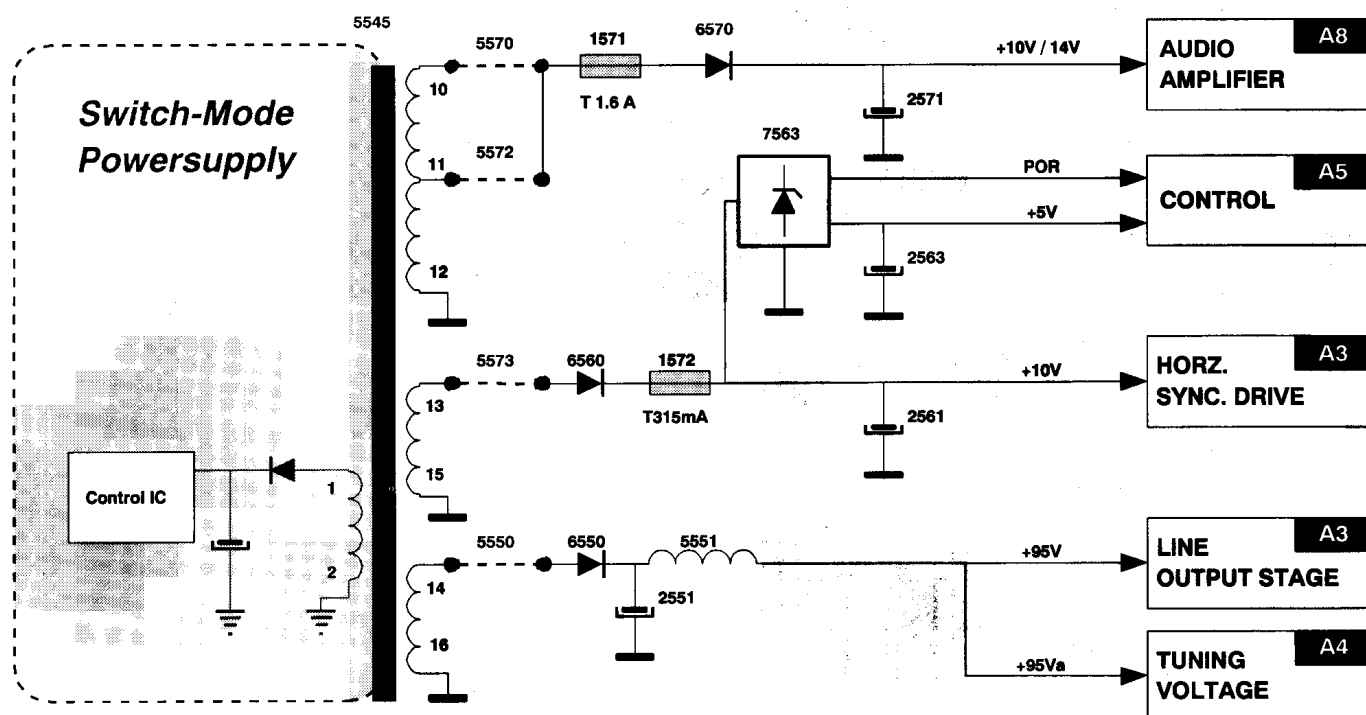
## D SCART (IF PRESENT)



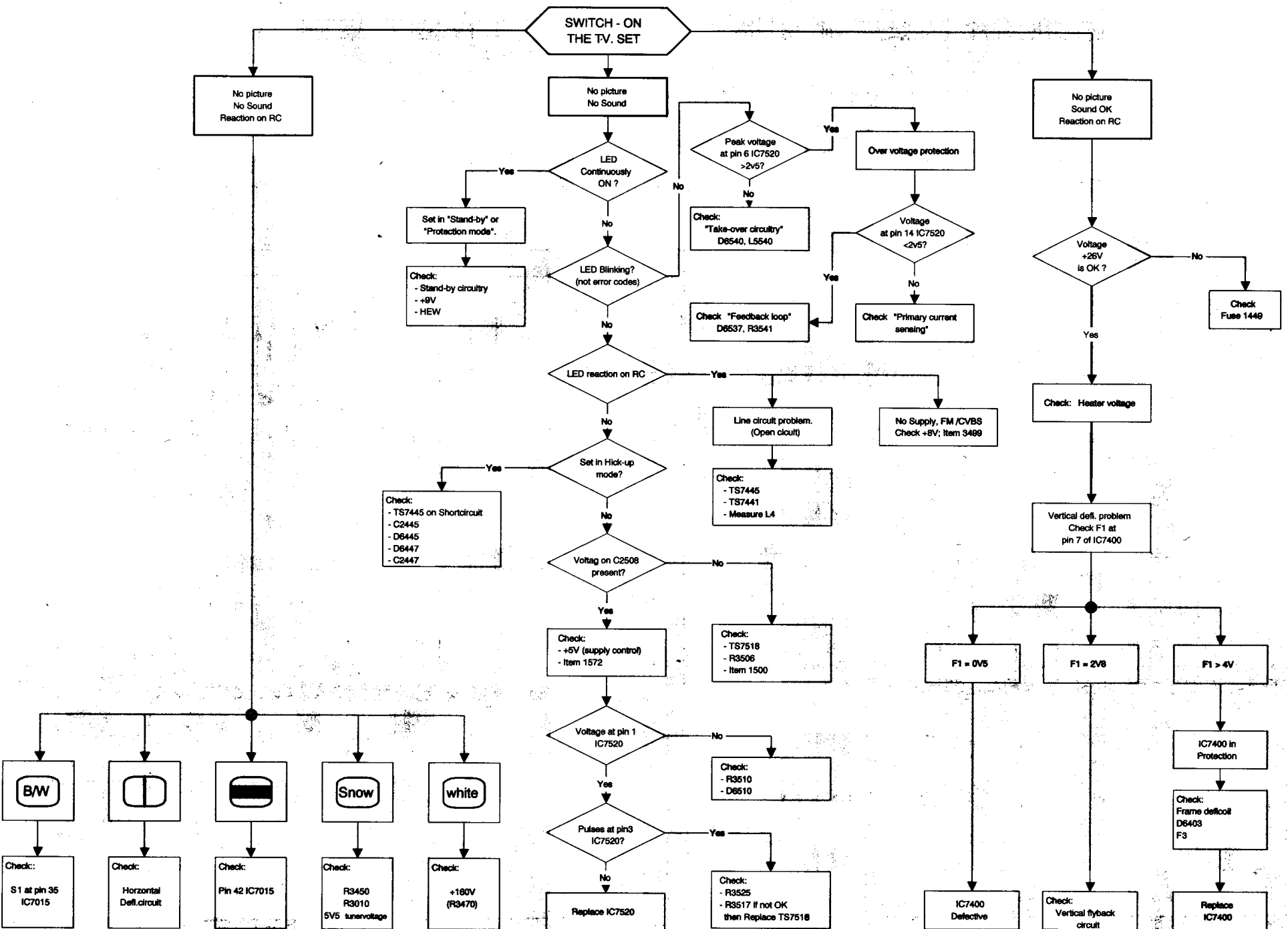
## A2 FRAME DEFLECTION







# 6. Fault finding tree & Repair facilities / Fehlersuchbaum & Reparaturhinweise / Aide au dépannage & Conseils pour la réparations



# Repair facilities

## 6.1 Functional blocks

On both the service printing on the copper and the component side, functional blocks are indicated by lines and text.

## 6.2 Test points

The L7.1 chassis is equipped with test points in the service printing on both sides of mono-board. These test points are referring to the functional blocks as mentioned above:

- \* P1-P2-P3, etc.: Test points for the power supply
- \* L1-L2-L3, etc.: Test points for the line drive and line output circuitry
- \* F1-F2-F3, etc.: Test points for the frame drive and frame output circuitry
- \* S1-S2-S3, etc.: Test points for the synchronization circuitry
- \* V1-V2-V3, etc.: Test points for the video processing circuitry
- \* A1-A2-A3, etc.: Test points for the audio processing circuitry
- \* C1-C2-C3, etc.: Test points for the control circuitry
- \* T1-T2-T3, etc.: Test points for the teletext processing circuitry

The numbering is done in a for diagnostics logical sequence; always start diagnosing within a functional block, in the sequence of the relevant test points, for that functional block.

## 6.3 Service mode

The service mode can be split into two parts: Service Default Mode (SDM) and Service Alignment Mode (SAM). For L7.1 these modes will be replaced by a combined mode, called SDAM.

The control system offers some features, which can be used by the service.

To entry the Service mode you have two possibilities:

- SDAM entry by Dealer Service Tool
- Short-circuit service pins M24 and M25 on PCB and switch power-on.

To leave the Service mode push the stand-by button; the error buffer will be cleared !!

Features are:

- Service settings after entry
- Service (sub)menu selection
- Error buffer display
- Software version & identification display
- Life timer (run timer) display

### 6.3.1 The initial state after switching on in service mode is:

System:

- For Multi-Europe sets PAL-BG
- For Multi-France sets SECAM-L
- For Bi-Norma and Tri-Norma sets PAL-M

Tuning:

- For sets with VST tuning:  
Programme number 1 is selected and the system will be tuned at the tuning data (for programme 1) read from EEPROM
- For sets with PLL tuning:  
Tune to a frequency of 475.25 MHz.

Further settings:

- The automatic switch off (no IDENT) timer and the sleep timer will be ignored.
- The child lock will be disabled.
- If the TV set was in hotel mode, this mode is disabled as long as the TV is in service mode.
- Brightness, saturation, sharpness, contrast and balance are initialised on 50% level.
- The volume is set to 25% level.
- After initialisation the TV set is normally controllable.
- To indicate that the TV is in service mode an "S" will be displayed (in green) in the top right corner of the screen. All other OSD will be in red.
- All displayed text strings in service mode are in English.
- The TV set will remain in SDAM after switching of by main switch; with stand-by you will leave this mode.

### 6.3.2 Other features

#### RAM test

At every start up of the TV, a read after write test for the complete RAM will be performed. If this check fails, the appropriate error number will be written in the error buffer. The patterns will be chosen in such a way that every bit of all bytes, will be written high and low.

#### Life timer (run timer)

During the life time cycle of the TV set a life timer is kept. This life timer only counts the normal operation hours, not the stand-by hours. Also at every switch on the life timer is incremented by one.

#### Error buffer

The last five errors, remembered from the EEPROM, are shown in the service main menu. This is called the error buffer. An error will be added to the buffer if this error differs from the last error in the buffer. The last found error is displayed on the left.

**Example:** Suppose the display shows:

3 4 1 3 1. This means the last found error is error number 3; the last found error but one is error number 4, and so on.

30000

43000

34300

## 6.4 Error codes

The following error numbers have been defined:

0 = No error

1 = Internal RAM error

2 = General I<sup>2</sup>C error

3 = EEPROM Configuration error (Checksum error)

4 = I2C error (TDA9840 / TDA9852)

5 = I2C error (TDA8374/75) (NOT IN L7.1)

6 = EEPROM error

7 = I<sup>2</sup>C error (PLL tuner)

## Repair facilities

### 6.5 SDAM mode

This menu is being displayed whenever SDAM is entered. In this menu the error buffer can be inspected, and the option byte(s) can be (re)programmed. The overview of the menu is shown below:

#### Explanation:

02031	The hexadecimal representation of the option byte contents.
3427	The hexadecimal value of the life timer.
2.2.1	The software identification, version and cluster.
S	The character "S" to indicate that the TV set is in service mode.
OP	A two character short name for the option to be selected.
VALUE	The value of the selected option.

OPTION CODE	OPERATION	SOFTWARE	S
02031	HOURS	VERSION	
	3427	2.2.1	
ERROR		34300	
OP		VALUE	

The MENU UP/DOWN command can be used to select the next/previous option; the MENU LEFT/RIGHT command can be used to change the option value.

The possible options are listed in the following table:

Table: Options description for L7 versions

#### Europe version

Full option name	Option name abbreviation	Value range	Available for
Virgin mode	VI	0 = off, 1 = on	Asian Pacific, Latin America, USA Bit 7 of byte 0
Hotel mode	HO	0 = not present, 1 = present	Asian Pacific PAL, Latin America Bit 6 of byte 0
Volume status	VS	0 = stored for all, 1 = stored per channel	Asian Pacific PAL Bit 5 of byte 0
Child lock	CL	0 = not present, 1 = present	Asian Pacific, Latin America, USA Bit 4 of byte 0
Hue	HU	0 = not present, 1 = present	Asian Pacific PAL Bit 3 of byte 0
AV source	AV	0 = not present, 1 = present	Asian Pacific, Latin America, USA Bit 2 of byte 0
UHF only	UH	0 = not present, 1 = present	Asian Pacific PAL Bit 1 of byte 0
Smart sound	SS	0 = not present, 1 = present	Asian Pacific PAL Bit 0 of byte 0
Smart picture	SP	0 = not present, 1 = present	Asian Pacific PAL Bit 7 of byte 1
Auto scan	AS	0 = not present, 1 = present	Asian Pacific, Latin America, USA Bit 6 of byte 1
60/80 programmes	PR	0 = 60 programmes, 1 = 80 programmes	Asian Pacific PAL Bit 5 of byte 1
Magnavox	MV	0 = not Magnavox, 1 = Magnavox	Asian Pacific PAL Bit 4 of byte 1
National brand	NB	0 = not National brand, 1 = National brand	Asian Pacific PAL Bit 3 of byte 1
Europe	EU	0 = not Europe, 1 = Europe	Asian Pacific PAL Bit 2 of byte 1
System	SY	0 = Single system (AP PAL, LatAm Tri-Norma), 1 = LA_BINORMA (LatAm Tri-Norma), 2 = LA_TRINORMA (LatAm Tri-Norma), 3 = AP-Multi, 4 = AP-Dual	Asian Pacific PAL, Latin America Tri-Norma Byte 2 is 0000 Byte 2 is 0001  Byte 2 is 0010 Byte 2 is 0011 Byte 2 is 0100

## LATAM version

Full option name	Option name abbreviation	Value range	Available for
Virgin mode	VI	0 = off, 1 = on	Bit 7 of byte 0
Child lock	CL	0 = not present, 1 = present	Bit 6 of byte 0
AV source	AV	0 = not present, 1 = present	Bit 5 of byte 0
Manual skip	SK	0 = not present, 1 = present	Bit 4 of byte 0
Vol limiter	VL	0 = not present, 1 = present	Bit 3 of byte 0
Auto scan	AS	0 = not present, 1 = present	Bit 2 of byte 0
System	SY	0 = Single system (AP PAL, LatAm Tri-Norma), 1 = LA_BINORMA (LatAm Tri-Norma), 2 = LA_TRINORMA (LatAm)	Byte 2 = 0000 Byte 2 = 0001 Byte 2 = 0010

## USA version

Full option name	Option name abbreviation	Value range	Available for
Virgin mode	VI	0 = off, 1 = on	Bit 7 of byte 0
Child lock	CL	0 = not present, 1 = present	Bit 6 of byte 0
Wake timer	WU	0 = not present, 1 = present	Bit 5 of byte 0
AV (ext)	AV	0 = not present, 1 = present	Bit 4 of byte 0
Vol limiter	VL	0 = not present, 1 = present	Bit 3 of byte 0
Auto scan	AS	0 = not present, 1 = present	Bit 2 of byte 0
Auto Cable detect	AC	0 = disable ,1 = enable	Bit 1 of byte 0

## LATAM close caption

Full option name	Option name abbreviation	Value range	Available for
Virgin mode	VI	0 = off, 1 = on	Bit 7 of byte 0
Child lock	CL	0 = not present, 1 = present	Bit 6 of byte 0
AV source	AV	0 = not present, 1 = present	Bit 5 of byte 0
Manual skip	SK	0 = not present, 1 = present	Bit 4 of byte 0
Vol limiter	VL	0 = not present, 1 = present	Bit 3 of byte 0
Auto scan	AS	0 = not present, 1 = present	Bit 2 of byte 0

## NTSC-AP

Full option name	Option name abbreviation	Value range	Available for
Virgin mode	VI	0 = off, 1 = on	Bit 7 of byte 0
Child lock	CL	0 = not present, 1 = present	Bit 6 of byte 0
AV source	AV	0 = not present, 1 = present	Bit 5 of byte 0
Auto scan	AS	0 = disable, 1 = enable	Bit 4 of byte 0
Auto Cable detect	AC	0 = disable, 1 = enable	Bit 3 of byte 0

The format of the option-code is the following:

7 6 5 4 3 2 1 0	7 6 5 4 3 2 1 0	7 6 5 4
x x x x x x x x	x x x x x x x x	x x x x
byte 0	byte 1	byte 2

All option-codes are presented hexadecimal in the service mode and not used bits are always 0.

**Example:** Option code C 0 1 0 4 in an Europe set means:  
binary 1100 0000 0001 0000 0100

This is a set with the following configuration:

- Virgin mode on
- Hotel mode present
- Magnavox set
- System PAL-I / PAL DK

If the EEPROM is replaced by a new one the set has to be installed according the option code.

## 6.6 Dealer remote used as a Dealer Service Tool (DST)

The purpose of the dealer remote is to enter the Service Alignment Mode or the Service Default Mode of the L7 chassis, simply by pressing respectively the ALIGN or the DEFAULT key of the DST.

DEFAULT key of the DST.

The main features are:

- Entering the dealer mode and executing commands in this mode must be done by RC5 remote control.
- Entry of the dealer mode is possible in all states, except from stand-by.
- Read the error buffer even if the OSD is not working at all. This is done via the blinking LED procedure (see 6.6).
- All software is suspended till the dealer remote mode is left.

The dealer mode is left if:

- The stand-by command is received

## 6.7 Blinking LED procedure

Via the DIAGNOSE 1 (for error 1) through the DIAGNOSE 5 (for error 5) commands of the DST, the error buffer can be made visible via the blinking LED. This is useful if the screen is not working properly.

The method is to use the LED pulses with as many pulses as the error number, followed by a time period of 3 seconds in which the LED is off.

E.g. error code 4 will result in four times the sequence LED on for 0.25 seconds / LED off for 0.25 seconds. After this sequence the LED will be off for 3 seconds.

## 6.8 Downloading of tuning data with the DST

Downloading of tuning data (programme number, frequency and system) via the DST will be made possible. This downloading is only possible in the version containing PLL tuning for Europe.

## 6.9 Hotel-mode and the hospital mode

The L7 chassis has one special mode, called the hotel mode.

Hotel mode:

- Installation menu cannot be entered.
- When entering the hotel mode the maximum volume will be the current value.
- The set will always switch to a selectable channel when the set is switched on.

Entering the hotel-mode:

- Select channel 38
- Push the menu button on the local keyboard and the OSD-button of the RC simultaneously for 3 seconds.

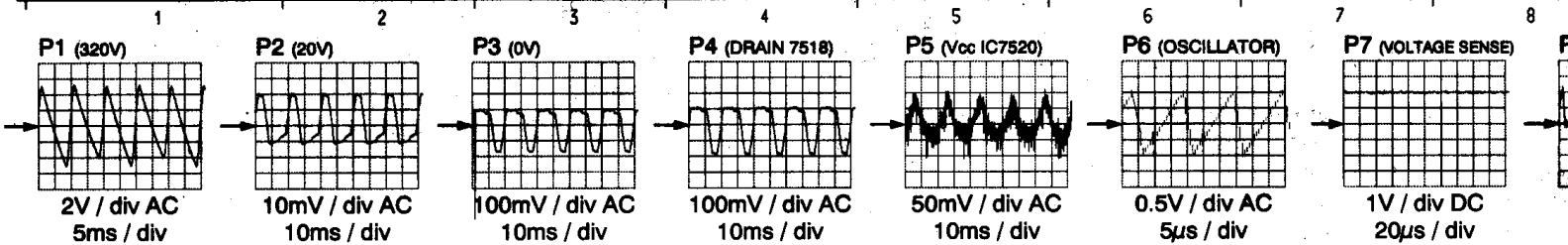
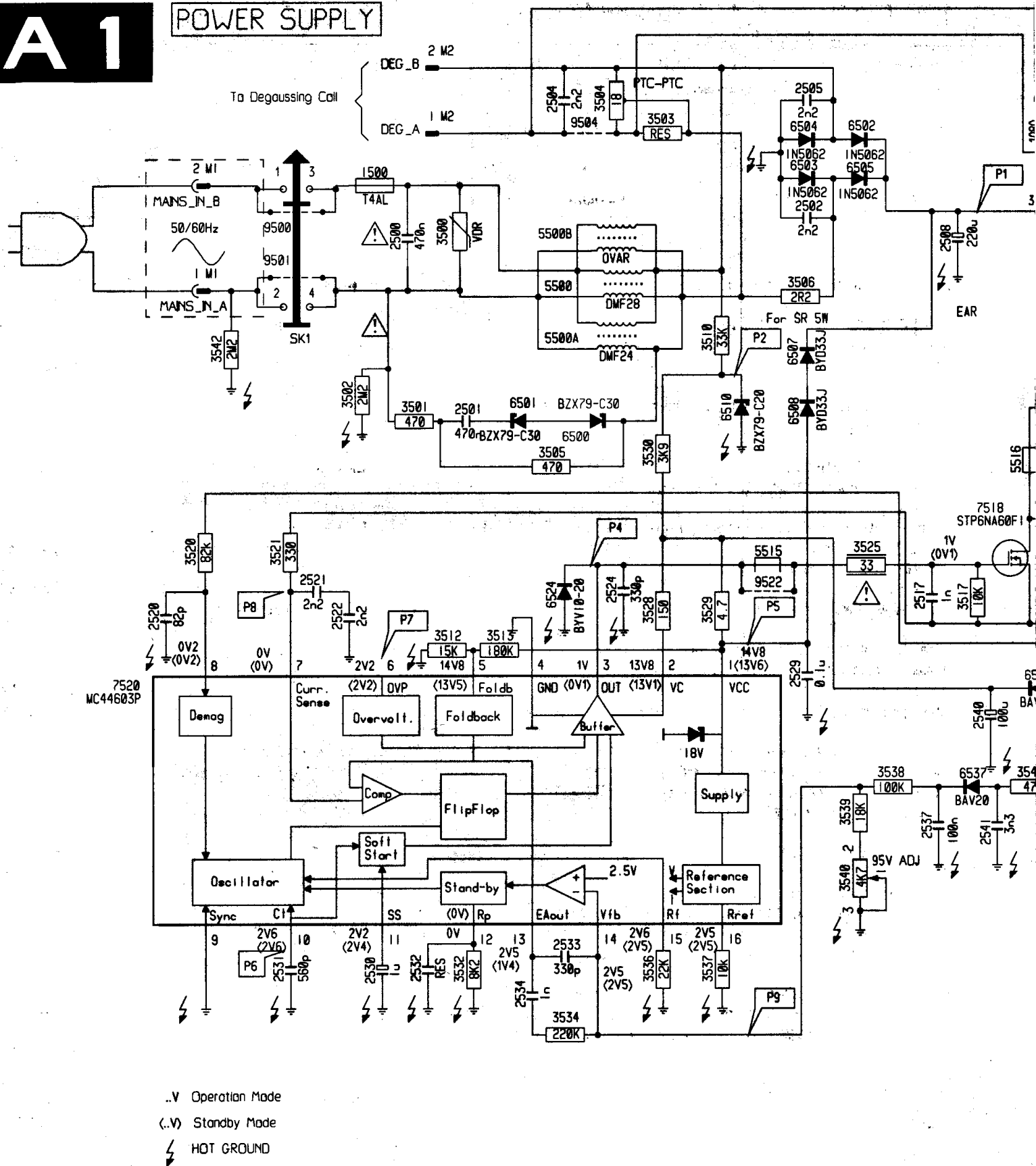
Leaving the hotel mode:

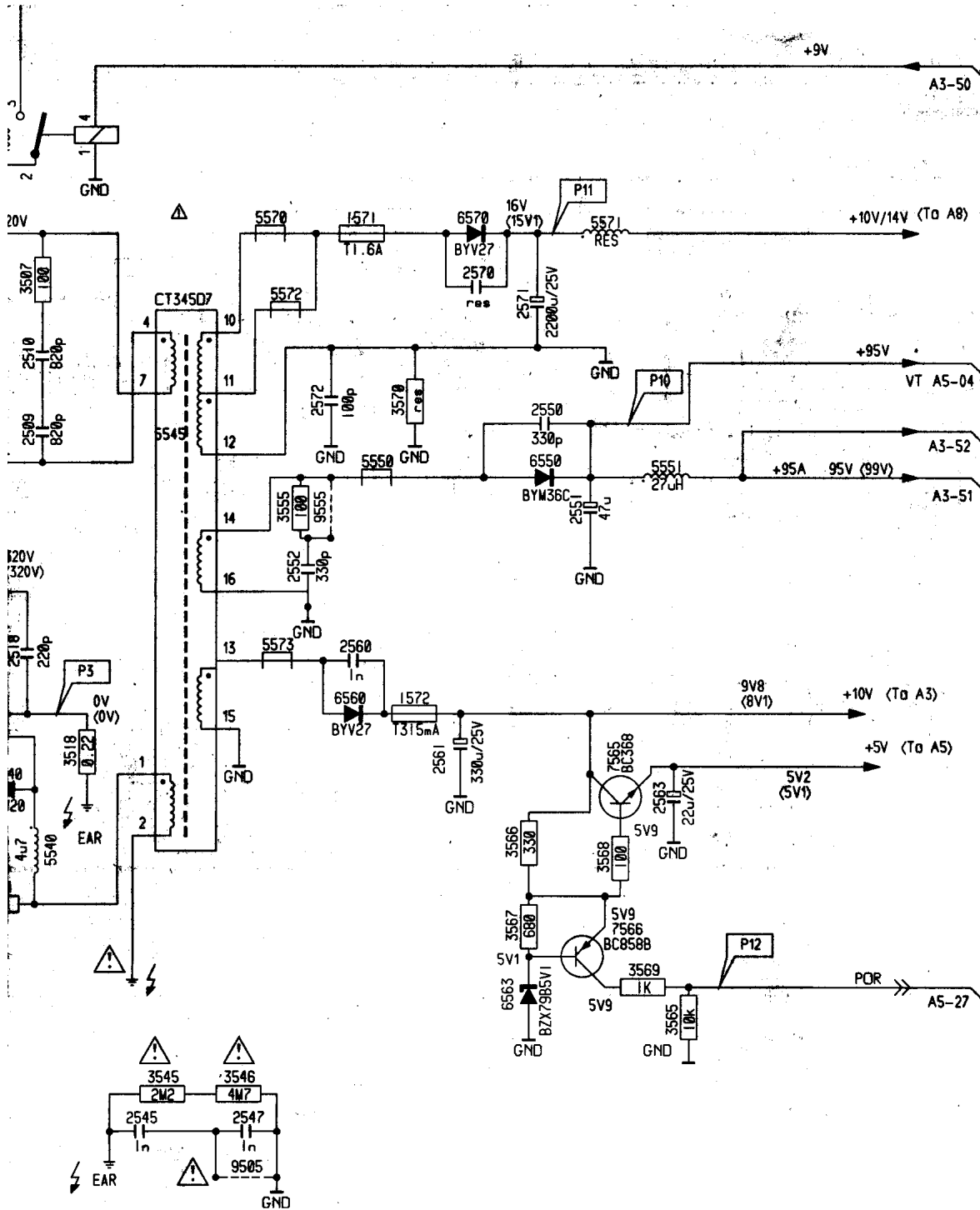
- Same as entering the hotel mode.

OSD will tell if hotel mode is on or off.

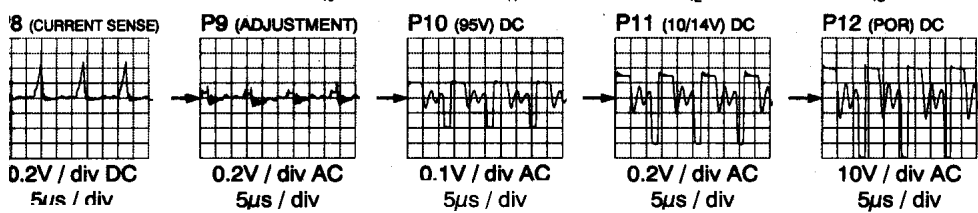
**A1**

**POWER SUPPLY**





1080	B 8	6550	D12
1500	B 3	6560	E10
1571	B11	6563	G12
1572	E11	6570	B11
2500	B 3	7518	D 8
2501	D 4	7520	F 1
2502	B 6	7565	F12
2504	A 5	7566	G12
2505	A 6	9500	B 2
2508	B 8	9501	C 5
2509	C 8	9504	A 5
2510	C 8	9505	H10
2517	F 7	9522	E 6
2518	F 8	9555	D10
2520	F 2	M1	C 2
2521	F 3	M1	B 2
2522	F 5	M2	A 4
2524	F 5	M2	A 4
2529	F 6	SK1	C 3
2530	H 3		
2531	H 2		
2532	H 4		
2533	H 5		
2534	H 4		
2537	G 7		
2540	F 8		
2541	H 9		
2545	H10		
2547	C12		
2550	D12		
2551	E11		
2552	D10		
2560	F11		
2561	F13		
2563	B11		
2570	C12		
2571	C10		
2572	B 4		
3500	A 4		
3501	A 4		
3502	D 5		
3503	A 5		
3504	A 5		
3505	D 5		
3506	C 6		
3507	B 8		
3510	C 6		
3512	F 4		
3513	F 4		
3517	F 9		
3518	F 9		
3520	F 2		
3521	F 7		
3525	F 5		
3528	F 5		
3529	D 5		
3530	D 5		
3532	H 4		
3534	H 5		
3536	H 5		
3537	H 6		
3538	F 7		
3539	G 7		
3540	G 7		
3541	F 8		
3542	C 2		
3545	H 9		
3546	H10		
3555	D10		
3556	D13		
3557	F12		
3566	G12		
3567	F12		
3568	G12		
3569	G12		
3570	C11		
5500	C15		
5500A	B 5		
5500B	B 5		
5515	F 6		
5516	D 8		
5540	F 9		
5545	C 9		
5550	D11		
5551	D13		
5570	B10		
5571	B12		
5572	C10		
5573	E10		
6500	D 5		
6501	D 4		
6502	B 7		
6503	B 6		
6504	B 6		
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6540	F 8		



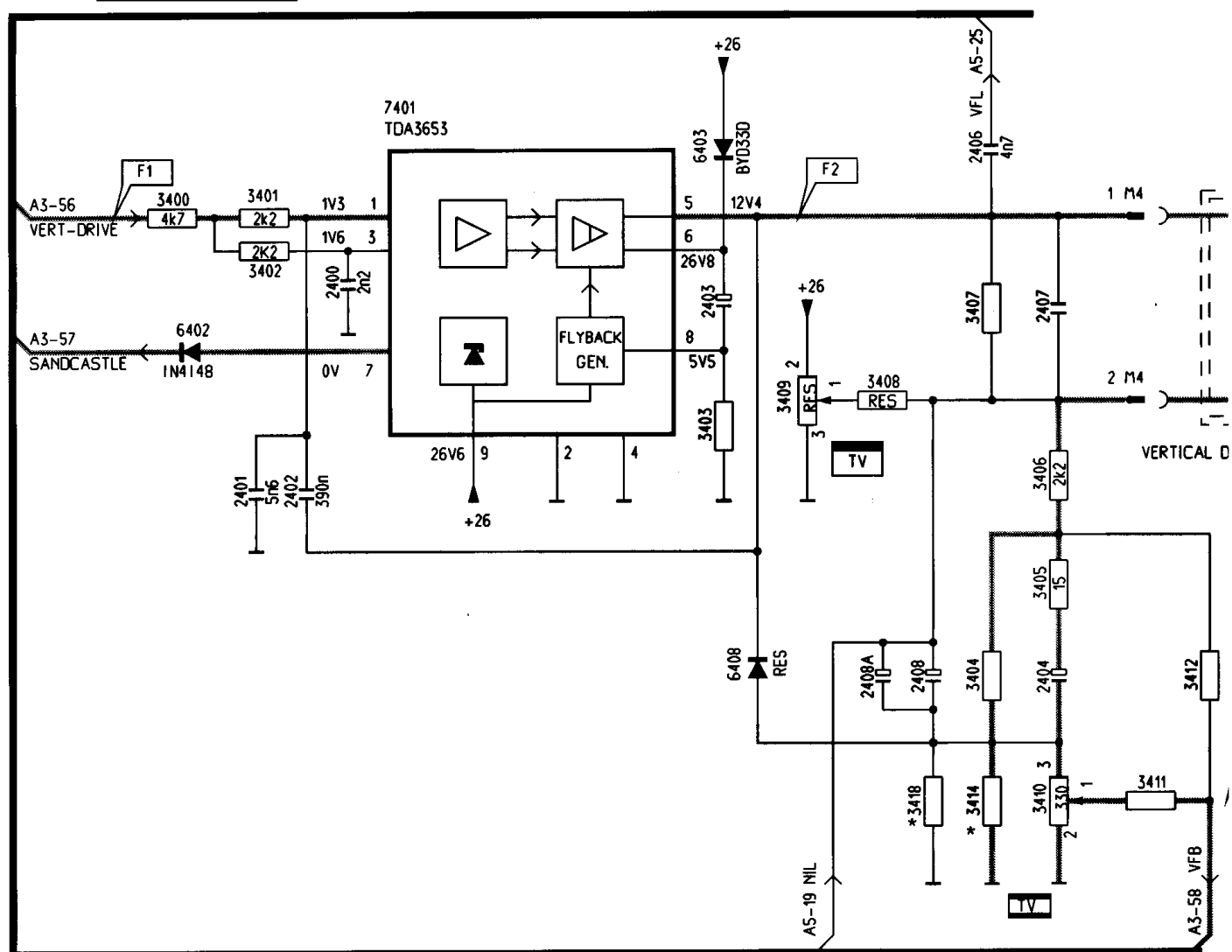
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080197

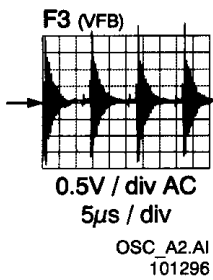
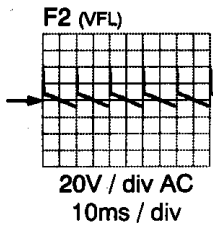
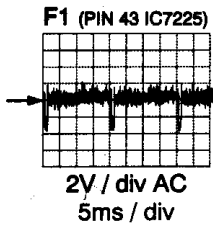
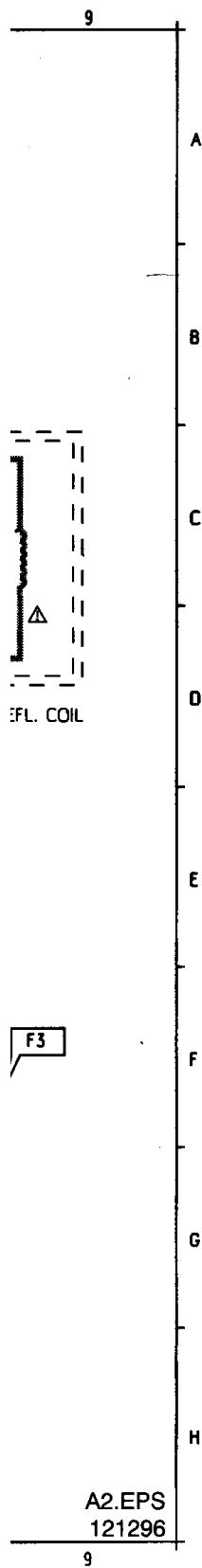


2400 C 3	2403 C 5	2407 C 7	3400 C 2	3403 D 5	3406 D 7	3409 D 6	3412 E 8	6402 C 2	7401 B 4
2401 D 3	2404 E 7	2408 E 7	3401 C 3	3404 E 7	3407 C 7	3410 F 7	3414 F 7	6403 B 5	M4 C 8
2402 D 3	2406 B 7	2408A E 6	3402 C 3	3405 E 7	3408 D 6	3411 F 8	3418 F 7	6408 E 6	M4 D 8

# A2

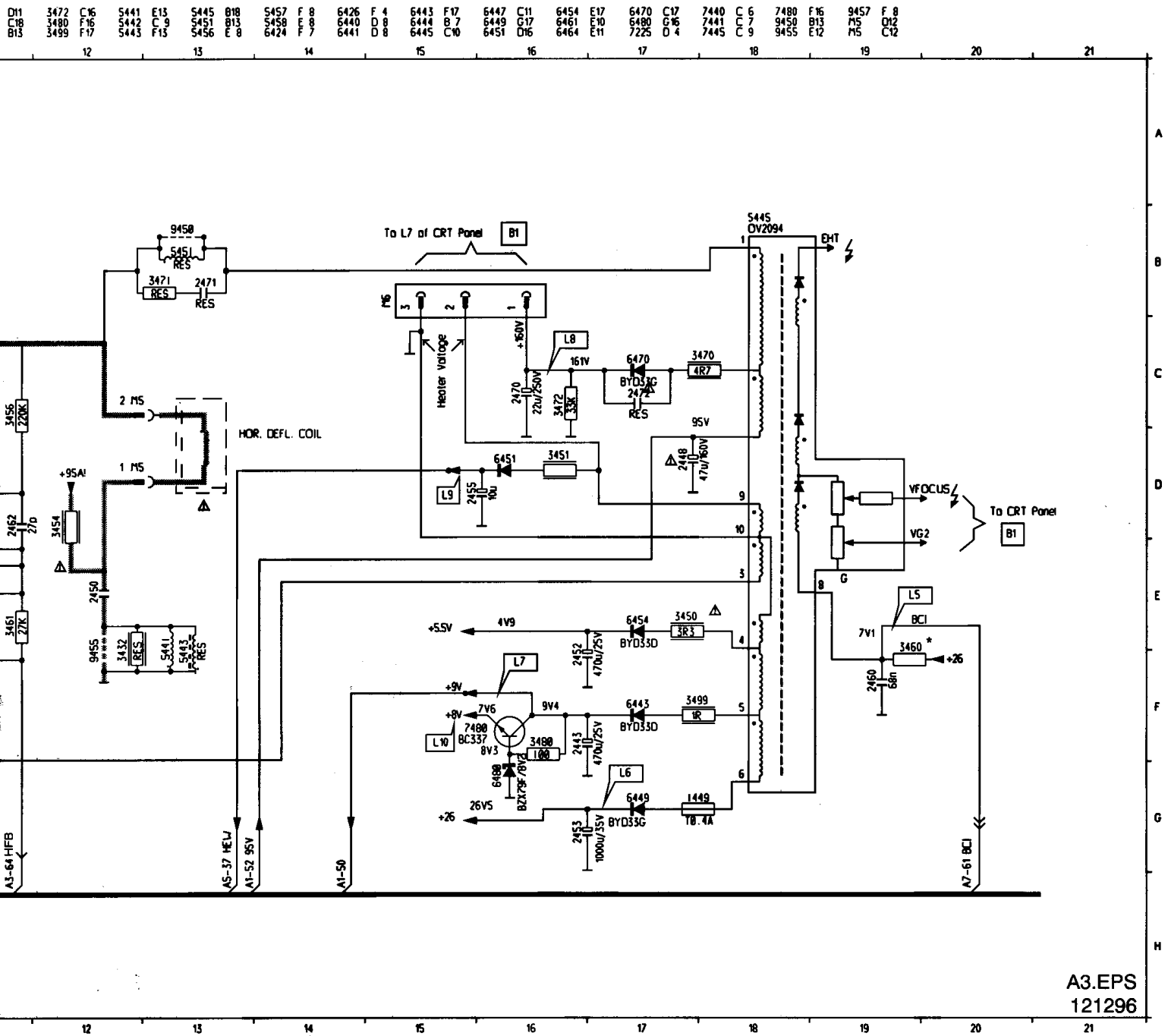
## FRAME-OUTPUT







# Synchronisierung + Ablenkung / Synchronisation + déviation



6 (26V) DC  
7 (9V) DC  
8 (160V) DC

S1 (SANDCASTLE)



2V / div AC  
50μs / div

OSC\_A3.A1  
080197

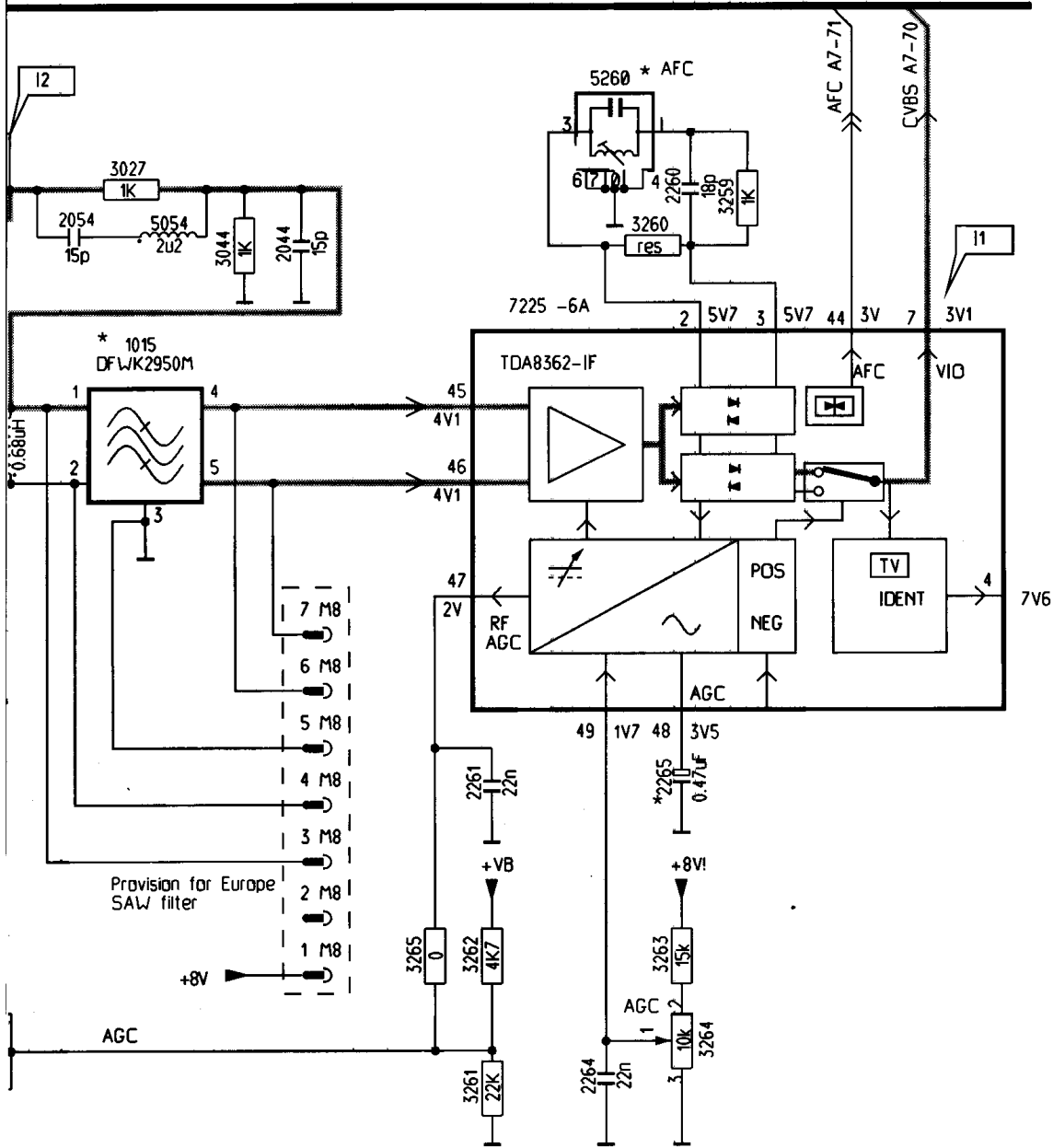
7

TUNER+IF

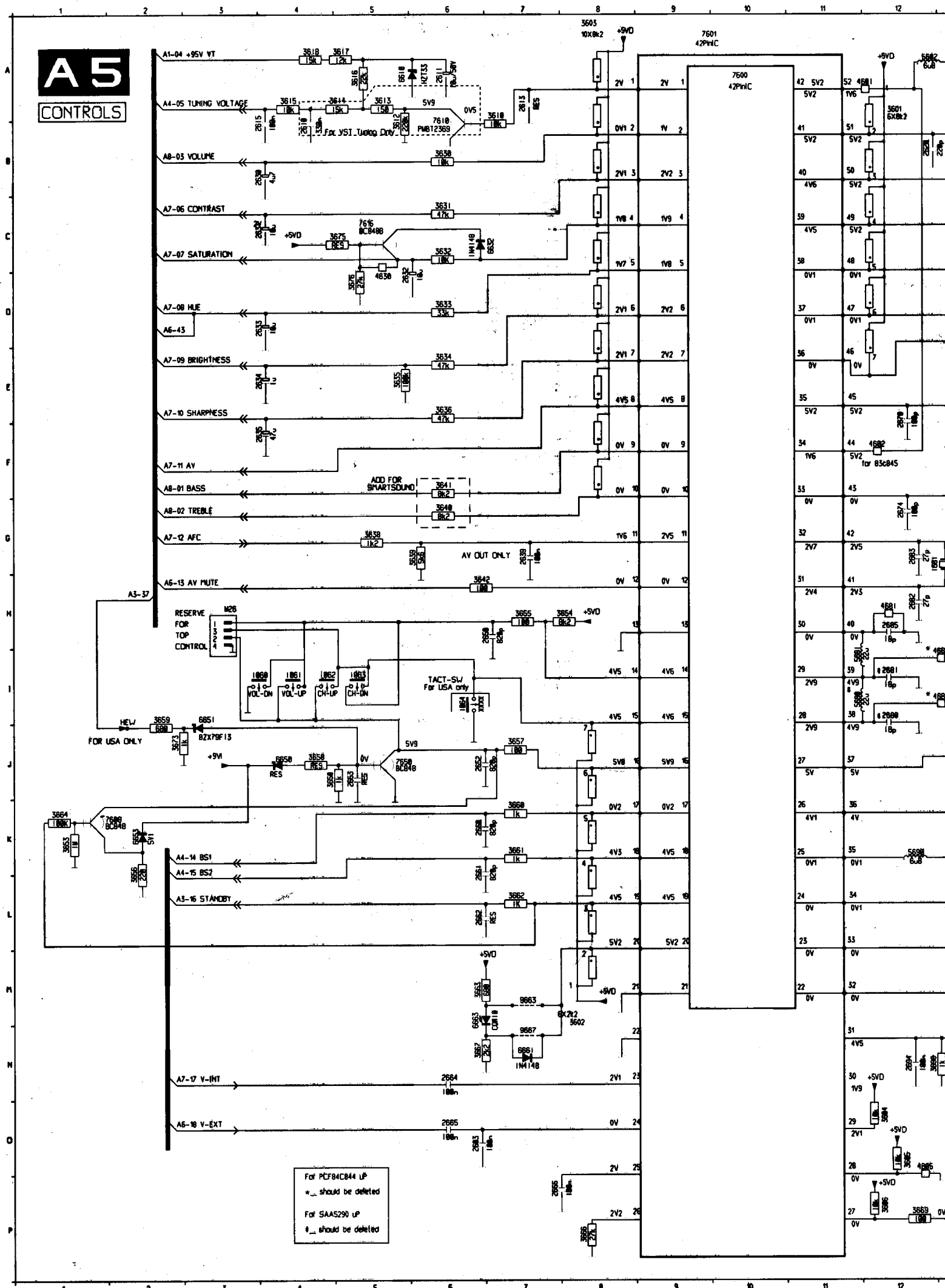


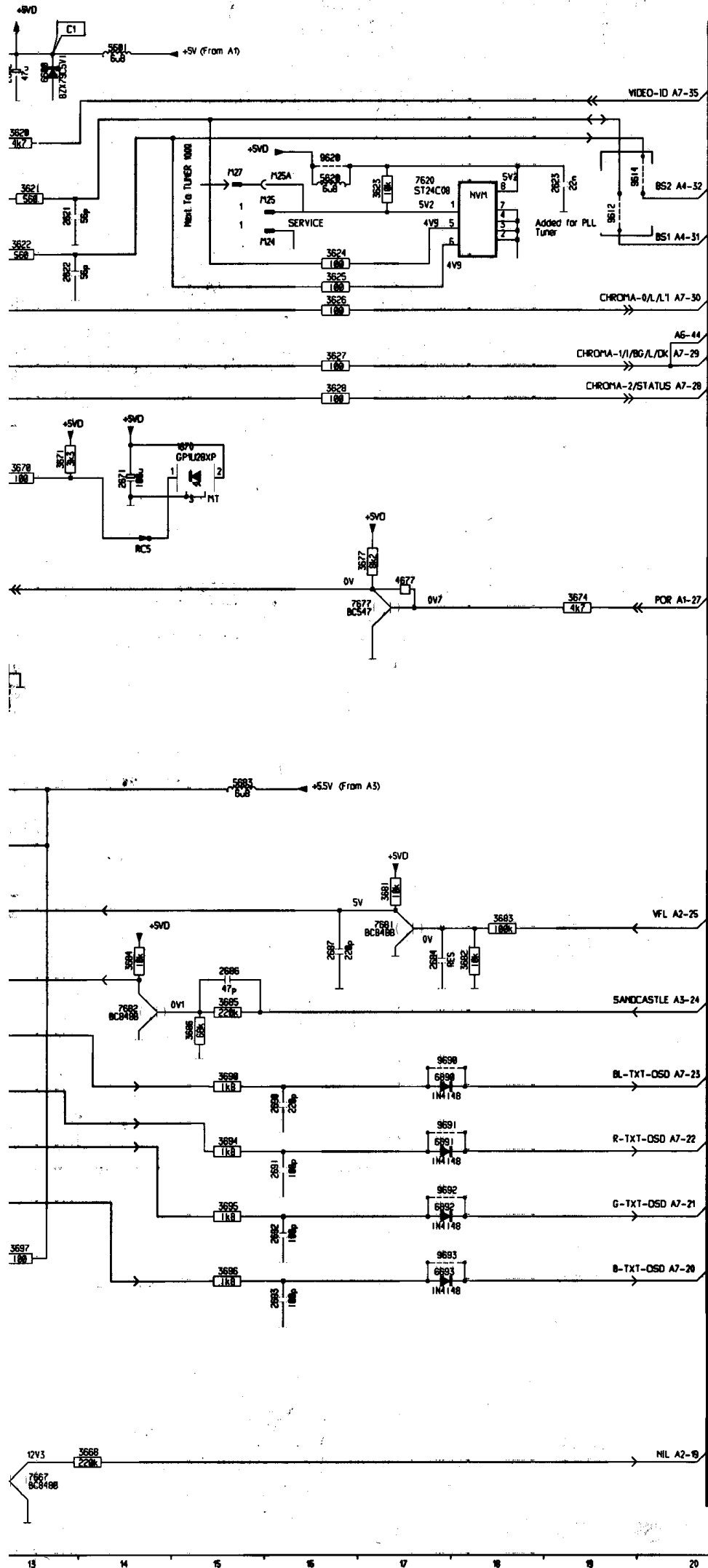
PCS 82 417

3264	G11	4011	D 2	5014	D 7	5260	B10	7002	F 3	7225	C10	9008	B 5	9011	B 5	M8	F 9	M8	E 9
3265	F 9	4013	E 2	5015	C 6	6016	F 7	7003	F 4	9003	E 5	9009	F 2	M8	F 9	M8	F 9	M8	E 9
4001	B 7	5002	B 4	5054	C 8	7001	E 2	7004	E 4	9004	E 5	9010	F 2	M8	F 9	M8	F 9	M8	E 9
	8				9				10		11			12				13	



**A5**  
CONTROLS

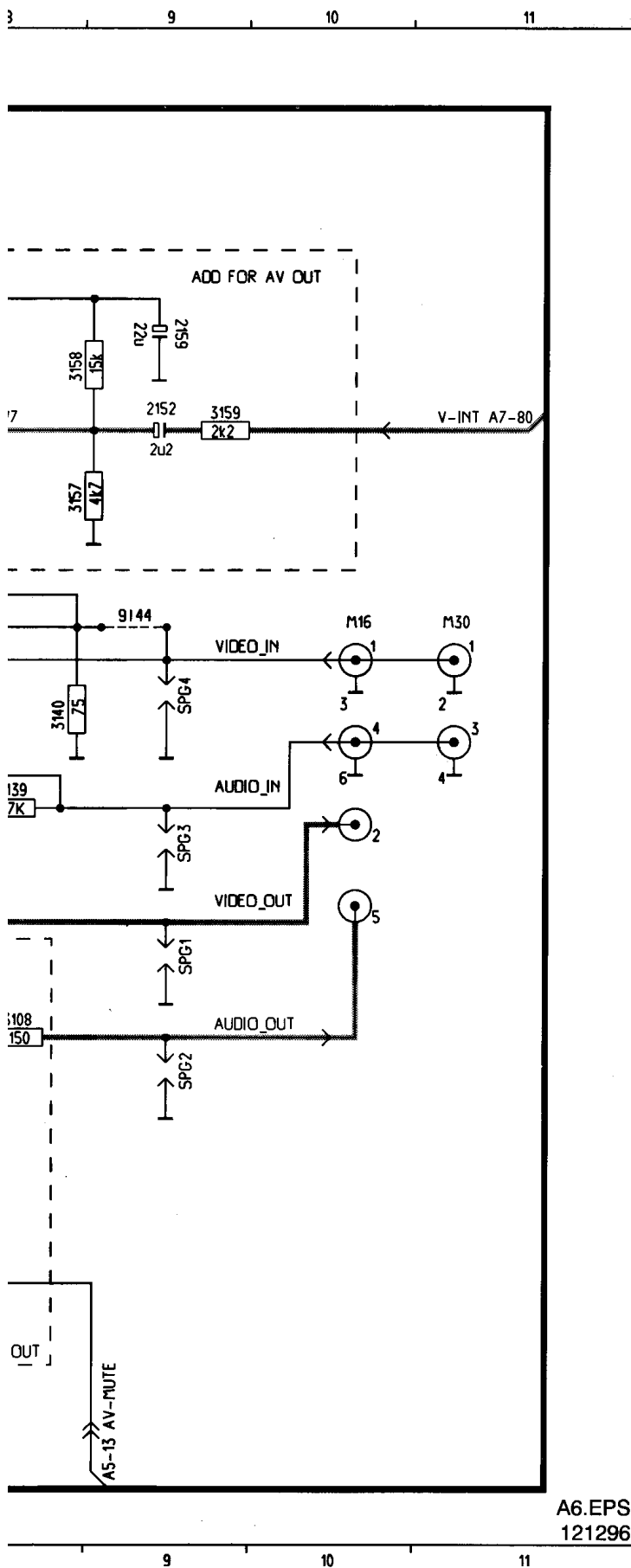






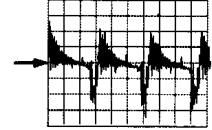


# AV entrée/sortie + FI son



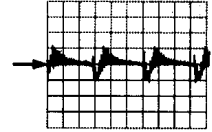
1101 C 3  
1102 C 3  
2101 F 3  
2102 F 3  
2103 G 5  
2104 F 3  
2107 I 3  
2108 B 5  
2109 G 6  
2138 F 4  
2150 F 7  
2151 G 8  
2152 C 9  
2159 B 9  
3101 G 6  
3102 F 7  
3104 C 5  
3105 D 5  
3106 E 2  
3107 B 4  
3108 F 8  
3109 F 7  
3110 G 7  
3111 E 2  
3112 G 4  
3113 H 4  
3116 D 2  
3118 C 5  
3119 H 7  
3120 H 6  
3138 E 8  
3139 E 8  
3140 E 8  
3144 D 8  
3150 C 7  
3155 C 7  
3157 C 8  
3158 B 8  
3159 C 9  
4205 D 6  
5100 B 7  
6102 C 4  
6144 E 7  
6151 C 7  
7102 D 4  
7109 G 7  
7110 H 6  
7150 C 8  
7225 D 2  
9101 C 5  
9102 C 5  
9144 D 9  
M16 D10  
M30 D11  
M34 F 7  
SPG1 F 9  
SPG2 G 9  
SPG3 G 9  
SPG4 D 9

A4 (OUTPUT)



50V / div AC  
20μs / div

A5 (INPUT)



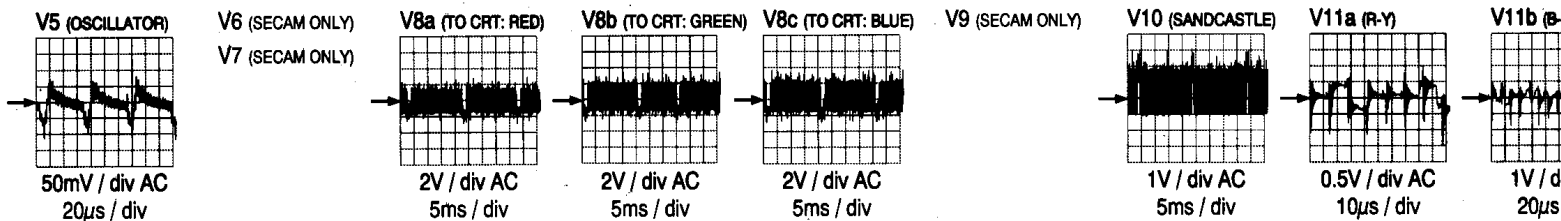
50V / div AC  
20μs / div

A8 (VCC) DC

OSC\_A6.AI  
121296

1	2	3	4	5	6	7	8	9	10	11
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## VIDEO & CHROMA PROCESSING



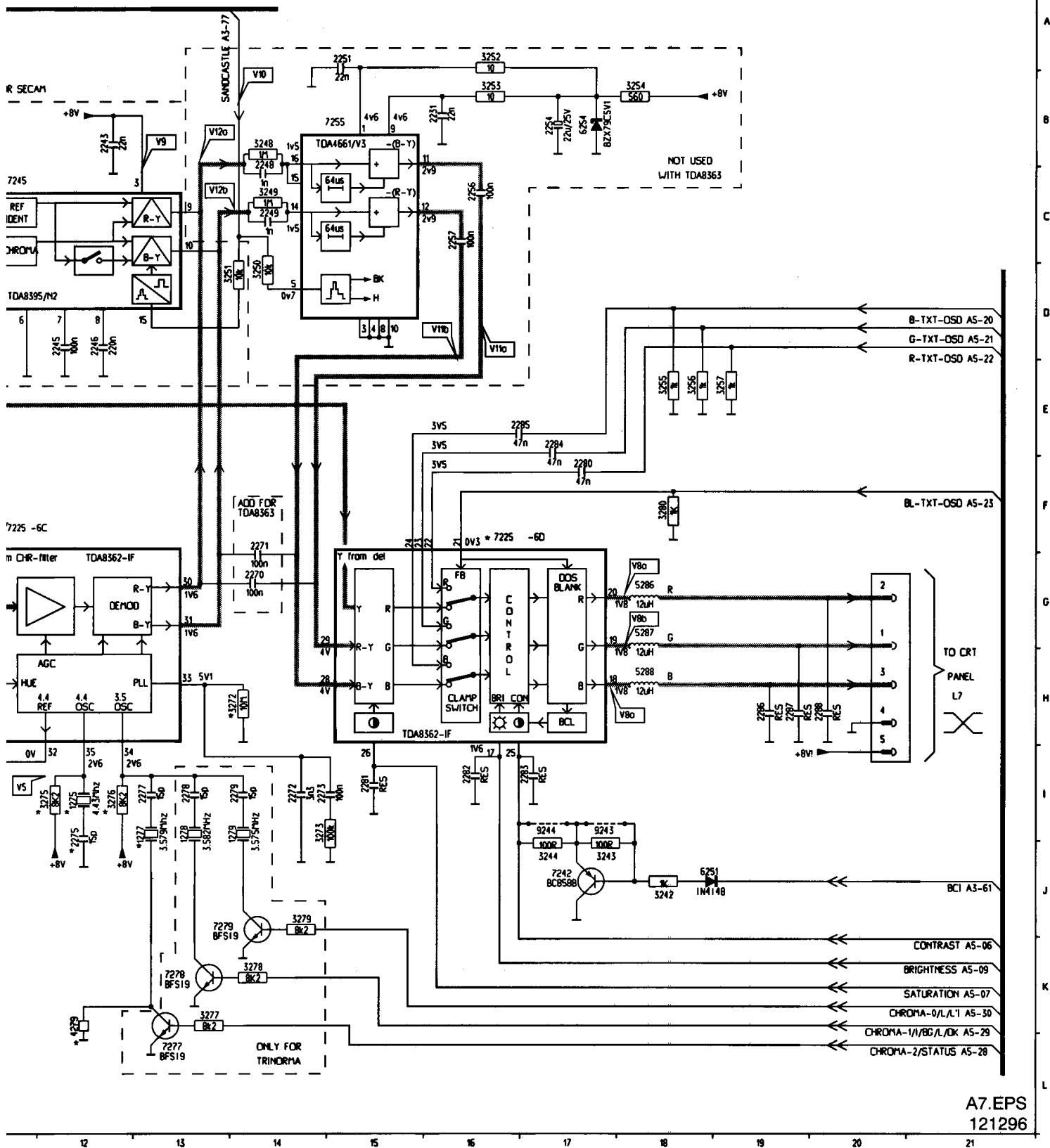
**V11b (B-**



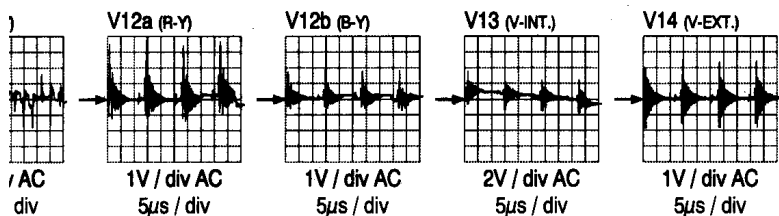
1V / d  
20μs

# Video & Chroma Verarbeitung / Vidéo & traitement chroma

3249 C14	3253 B16	3257 E19	3273 I14	3278 K14	4201 C.5	5209 F.5	6251 J18	7216 B.6	7225 F16	7255 B14	7279 J15	9246 I11
3250 D14	3254 B18	3270 I10	3275 I12	3279 J14	4279 K12	5286 G18	6254 B17	7217 B.7	7240 C.9	7269 I.9	9243 I17	
3251 D14	3255 E18	3271 I10	3276 I12	3280 F18	5207 C.5	5287 G18	7214 G.4	7225 F11	7242 J17	7277 L13	9244 I17	
3252 A16	3256 E18	3272 H14	3277 K13	4200 F.7	5207 C.5	5288 H18	7215 G.4	7225 D.7	7245 C11	7278 K13	9245 D11	



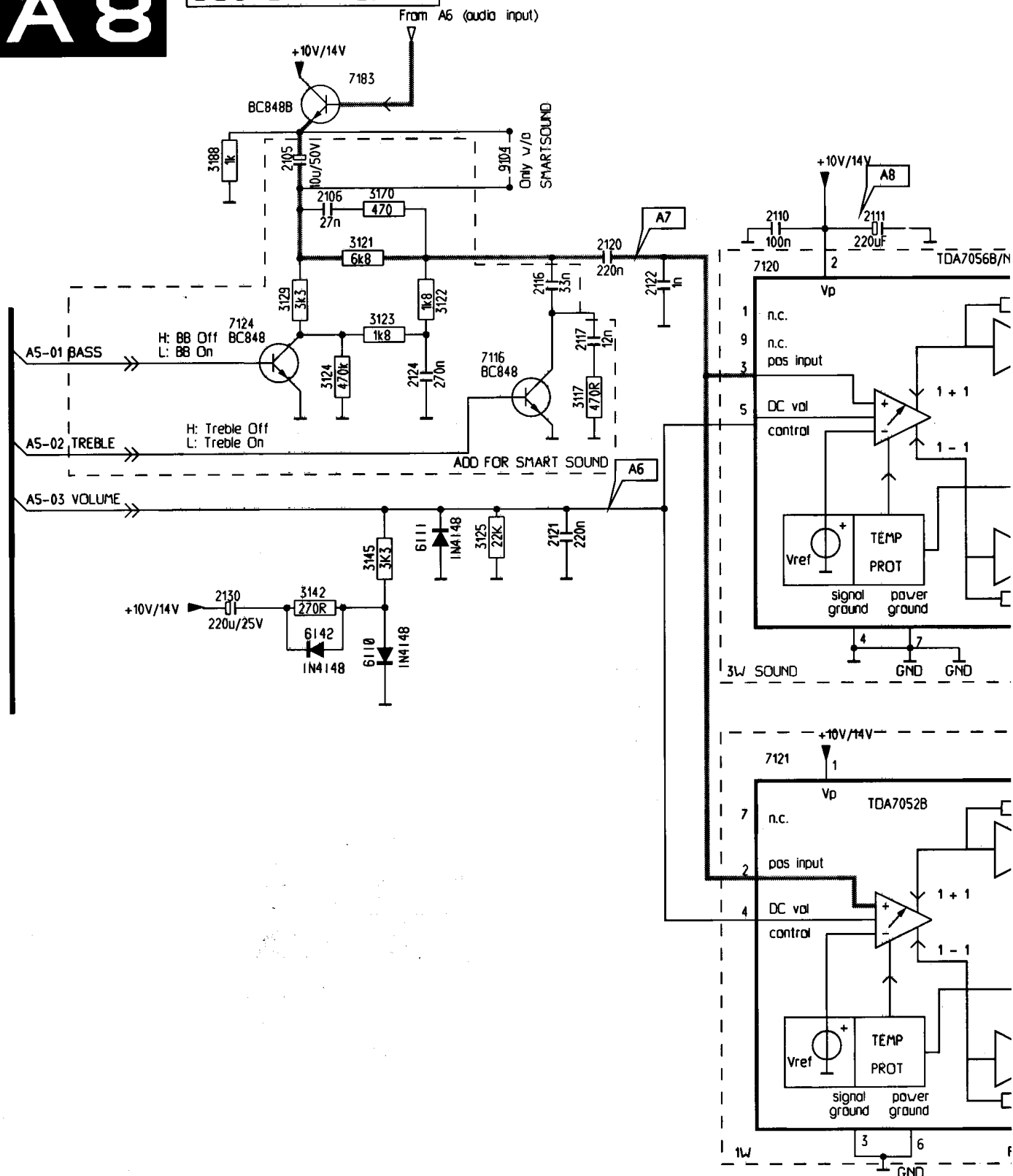
A7.EPS  
121296



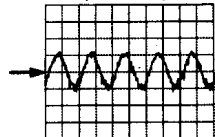
OSC A7.AI  
070897

# A8

## SOUND AMPLIFIER

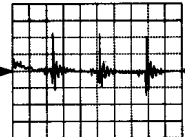


A1 (OUTPUT)



50mV / div AC  
0.5ms / div

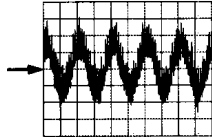
A2 (OUTPUT)



10mV / div AC  
20μs / div

A6 (VOLUME) DC

A7 (INPUT)

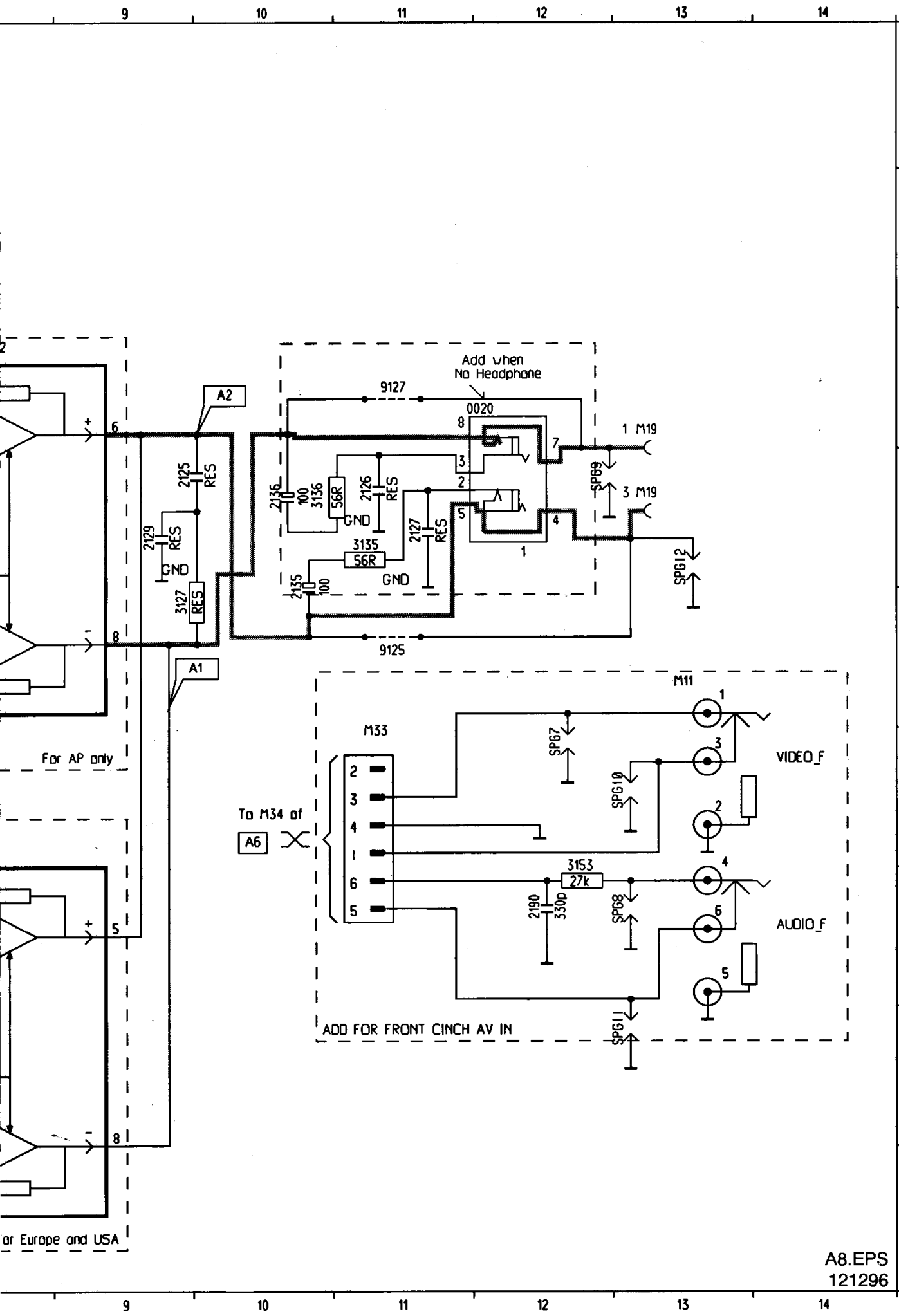


10mV / div AC  
0.5ms / div

A8 (VCC) DC

OSC\_A8.AI  
101296

Amplificateur son



0020	C11
2105	B 3
2106	B 3
2110	B 6
2111	B 7
2116	C 5
2117	C 5
2120	F 5
2121	F 5
2122	C 6
2124	D 4
2125	D 9
2126	D 11
2127	D 11
2129	D 9
2130	F 3
2135	D 10
2136	D 10
2190	G 12
3117	D 5
3121	C 3
3122	C 4
3123	C 4
3124	D 3
3125	F 4
3127	F 9
3129	C 3
3135	D 11
3136	D 10
3142	F 3
3145	F 4
3153	G 12
3170	B 4
3188	B 2
6110	F 4
6111	F 4
6142	F 3
7116	C 4
7120	C 6
7121	C 6
7124	C 3
7183	A 4
9104	B 4
9125	E 11
9127	C 11
M11	E 13
M19	C 13
M19	D 13
SPG10	F 13
SPG11	H 13
SPG12	D 13
SPG7	F 12
SPG8	G 13
SPG9	D 12

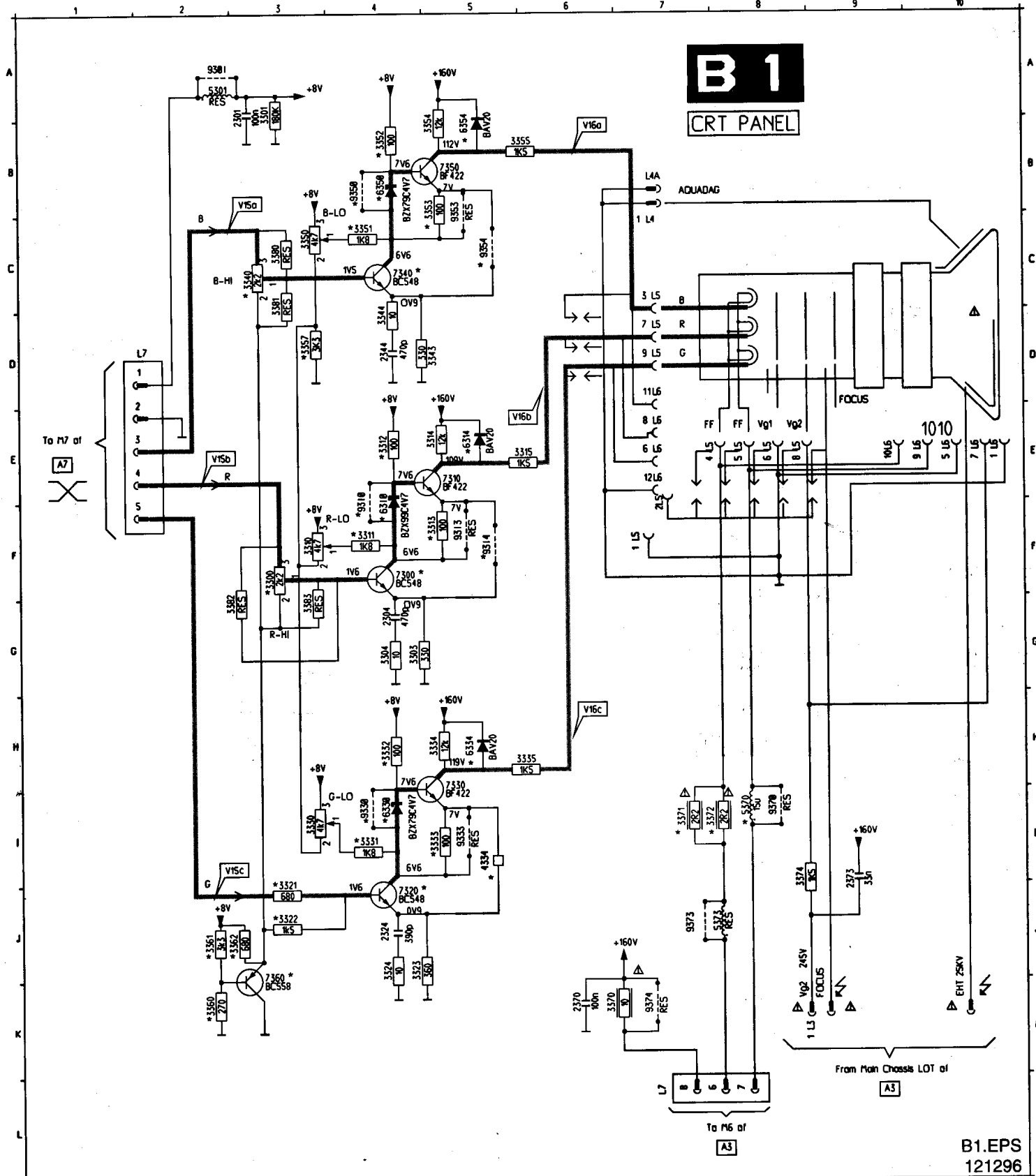
F

G

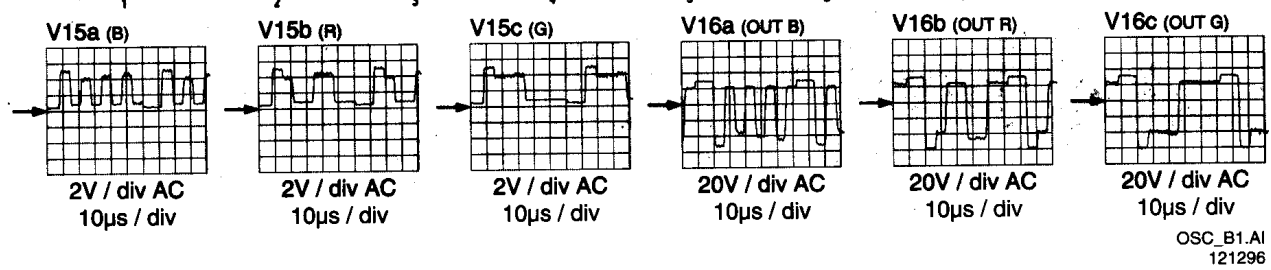
H

I

2301	A 3	3301	A 3	3314		3321	1 4	3344	4	3357	D	3374	8	3370	1 8	6354	B 5	7360	4 3	9350	B 4	L4	7	L5	D 7	L6	F 7
2304	A 4	3303	A 3	3315		3322	1 4	3345	4	3358	D	3375	8	3371	1 8	6355	B 5	7361	4 3	9351	B 4	L4A	7	L5	D 7	L6	F 7
2304	A 4	3304	A 3	3316		3323	1 4	3346	4	3359	D	3376	8	3372	1 8	6356	B 5	7362	4 3	9352	B 4	L4A	7	L5	D 7	L6	F 7
2344	D 4	3310		3327		3324	1 4	3347	4	3360	D	3377	8	3373	1 8	6357	B 5	7363	4 3	9353	B 4	L4A	7	L5	D 7	L6	F 7
2370	F 3	3311		3328		3325	1 4	3348	4	3361	D	3378	8	3374	1 8	6358	B 5	7364	4 3	9354	B 4	L4A	7	L5	D 7	L6	F 7
2373	F 3	3312		3329		3326	1 4	3349	4	3362	D	3379	8	3375	1 8	6359	B 5	7365	4 3	9355	B 4	L4A	7	L5	D 7	L6	F 7
3300		3313		3330		3327	1 4	3350	4	3363	D	3380	8	3376	1 8	6360	B 5	7366	4 3	9356	B 4	L4A	7	L5	D 7	L6	F 7



B1.EPS  
121296



## 8.1 Settings on the carrier panel

### 8.1.1 +95V supply voltage

Connect a multimeter (DC) across C2531. Set brightness at mid position and contrast at maximum. Apply a pattern generator with a colour bar. Adjust potentiometer **R3540** to  $+95V \pm 0.5V$  DC.

### 8.1.2 Horizontal centring

Is adjusted with potentiometer **R3420**.

### 8.1.3 Vertical centring

Can be adjusted with **R3409**.

### 8.1.4 Picture height

Is adjusted with potentiometer **R3410**.

### 8.1.5 Focusing

Is adjusted with the focusing potentiometer in the line output transformer 5445 (if necessary set brightness at minimum and contrast at maximum for focus adjustment).

### 8.1.6 RF AGC adjustment

Connect a pattern generator (e.g. PM5518) to the aerial input with RF signal amplitude = 1 mV. Connect a multimeter (DC) at pin 1 of tuner. Adjust **R3264** so that voltage at pin 1 of tuner is  $3.3 \pm 0.2V$  DC.

### 8.1.7 Picture demodulator adjustment

Connect a pattern generator (e.g. PM5518) with a cross hatch. Connect an oscilloscope (1ms/div) to pin 7 of IC7225-6A and adjust **L5260** so that the overshoot response is minimum, see Fig. 8.1.  
Select a colour bar signal and verify if the picture is all right.

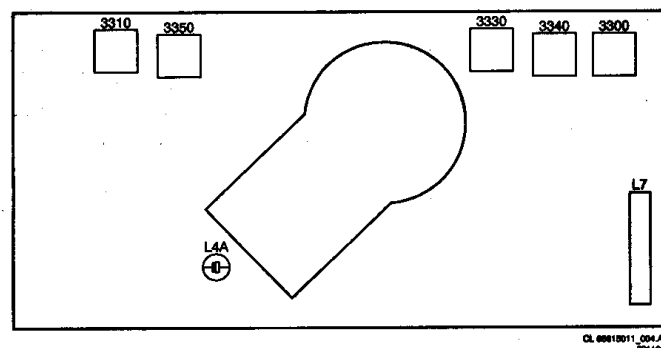


Fig. 8.1

## 8.2 Settings on the CRT panel

### 8.2.1 Vg2 cut off adjustment

Connect a pattern generator (e.g. PM5518) and set it to white raster pattern. Set contrast and the Vg2 potentiometer (in line output transformer) minimum. Adjust with brightness control the top video level at pin 4L7 to the same voltage level of the emitter of transistor 7360.

Pre-adjust the black level preset potentiometer **R3310** and **R3350** fully counter-clockwise. Adjust Vg2 potentiometer of LOT 5445 until green just becomes visible. Adjust the other two guns with their potentiometer: **R3350** for blue and **R3310** for red. All three colour shall give the same reading for a white picture.

### 8.2.2 White-D adjustment

Use the same signal as prescribed in 2.1. Adjust contrast to such a level that red is good visible.  
Adjust potentiometers **R3340** (B) and **R3300** (G) to have a correct White-D picture.



# 9. Circuit description new circuitries

Power supply (diagram A1)

## 9.1 Introduction

### 9.1.1 General

The L7 switched mode power supply (SMPS) is mains isolated. The control IC7520 (MC44603P) gives the pulses for driving FET 7518 with duty cycle control at a fixed frequency of nominal 70 kHz in normal operation (in standby, slow-start and overload situation the SMPS runs at other frequencies than these 70 kHz). This SMPS works with a switching FET, no opto-coupler and no thyristor switching windings on the secondary side.

IC7520 is featured with a slow-start circuitry and has over- and undervoltage-protection of the secondary supply voltages. Unload and overload (short-circuit) protection is also included. In case the load decreases under a certain threshold level the SMPS will switch into standby-mode (in standby the SMPS is in the so called "reduced frequency mode"; nominal 20 kHz).

The +VBATT output gives a stabilised +95V for 14" and +100V for 21" in normal operation and approx. 115V DC in standby for mode (the supply voltage +8V is "down", so the line output is shut "down").

### 9.1.2 Output voltages

- +10V / 14V for the audio amplifier
- +5V for the control part
- +10V for the horizontal synchronisation drive
- +95V for the line output stage

### 9.1.3 Duty cycle and T-on, T-off, T-dead

The duty cycle of the power supply depends on T-on of FET TS7518 which is controlled by pin 3 of IC7520. The IC detects the variations of the +VBATT (the secondary side of T5545) via sensing-winding 1-2 at the primary side of T5545. The switching period of FET 7518 is divided in three main areas; T-on, T-off and T-dead (see Fig. 9.1).

- During **T-on** FET 7518 conducts and so the energy which is extracted from the mains, is stored into the primary winding 4-7 of transformer T5545 with a linear increasing primary current (slope depends on the voltage across C2508). Via T-on regulation by pin 3 IC7520 the duty cycle of the SMPS and so the +VBATT is controlled.
- During **T-off** FET 7518 does not conduct and so all energy "inside" the transformer is supplied to the load via secondary windings of T5545 and the secondary diodes (D6550, D6560 and D6570). The current through the secondary side of the transformer decreases with a linear slope (slope depends on the voltage at the secondary side of T5545).
- During **T-dead** FET 7518 does not conduct and so no energy is extracted or supplied ( $I_{sec}$  is zero).

## 9.2 Primary side

### 9.2.1 Mains input and degaussing

**Mains voltage** is filtered by L5500, full wave rectified by a diode bridge and smoothed by C2508 to the DC input voltage for the SMPS at pin 7 of T5545 (e.g. 300V DC for 220V AC mains).

**Degaussing:** R3504 is a dual PTC (2 PTC's in one housing). After switching "on" the set, the PTC is cold so low-ohmic and so the degaussing current is very high. After degaussing, the PTC is heated, so high-ohmic, so in normal operation the degaussing current is very low.

### 9.2.2 Start up and take over

**Start-up:** Via the start-up circuitry R3530 and R3529 one side of the 220V AC mains is used to start-up IC7520 via the supply pin ( $V_{pin1}$ ). As long as  $V_{pin1}$  has not reached 14V5, IC7520 does not start up and only sinks 0.3 mA;

As soon as  $V_{pin1}$  reaches the 14V5, IC7520 starts (FET 7518 into conduction) and pin 1 sinks a typical supply current of 17 mA. This supply current can not be delivered by the start-up circuit, so a take-over circuit has to be available. If no take-over takes place, the voltage on pin 1 will decrease and IC7520 switches off. In that case the restart will start again. Note; This power supply is a SMPS (Switched Mode Power Supply) but not a SOPS (Self Oscillating Power Supply).

**Take over of IC7520:** During start-up a voltage across winding 1 - 2 is built up. At the moment the voltage across winding 1 - 2 reaches approx. +12V, D6540 starts conducting and takes over the supply voltage  $V_{pin1}$  of IC7520 (take over current is approx. 17 mA).

## 9.3 Control circuitry

### 9.3.1 IC7520 control mechanisms

IC7520 controls the T-on of FET 7518 in all operation modes by 3 mechanisms:

- "Secondary-output-voltage-sensing" controls the secondary output voltages (via the feedback voltage  $V_{pin14}$ ).
- "I-prim current sensing" controls both the secondary output voltages and the maximum I-prim (via the current sense voltage  $V_{pin7}$ ).
- "Demagnetisation control" prevents the transformer T5545 from going into saturation via the so called "DEMAG" function at pin 8 (this causes slow-start operation).

### 9.3.2 Secondary output voltages feedback (pin 14 of IC7520)

Winding 14 - 12 has the same polarity as the secondary windings which are supplying the load. During T-off the secondary windings and so winding 14 - 12 are positive. D6537 conducts and so charges C2537; the DC level across C2537 is a reference for the secondary output voltages (e.g. the +VBATT). Via R3538, R3539 and potentiometer R3540 (for adjusting the +VBATT) this DC-voltage is brought to the required level for the error amplifier in IC7520 at pin 14. This voltage  $V_{pin14}$  is called feedback voltage and is used to control the secondary output voltages.

### 9.3.3 I-prim sensing (pin 7 of IC7520)

The current sense voltage  $V_{pin7}$  is a measure for the I-prim through FET 7518. The I-prim is converted into a voltage by R3518. The current sense voltage  $V_{pin7}$  is used to control both the secondary output voltages and the maximum I-prim (see peak current limiting).

### 9.3.4 Demagnetisation control (via pin 8 of IC7520)

Winding 1 - 2 has the same polarity as the secondary windings which are supplying the load. As a result the voltage across this winding is negative during T-on, positive during T-off and oscillating during T-dead. The so called demagnetisation (block "DEMAG" in IC7520) function at pin 8 of IC7520 is used for blocking the output  $V_{pin3}$  during the time that there is still energy in the transformer ( $I_{sec}$  not zero).

This is realised by delaying the T-on until the demagnetisation is completely finished. In this way the currents and voltages at the moment of switching "on" the FET are controlled.

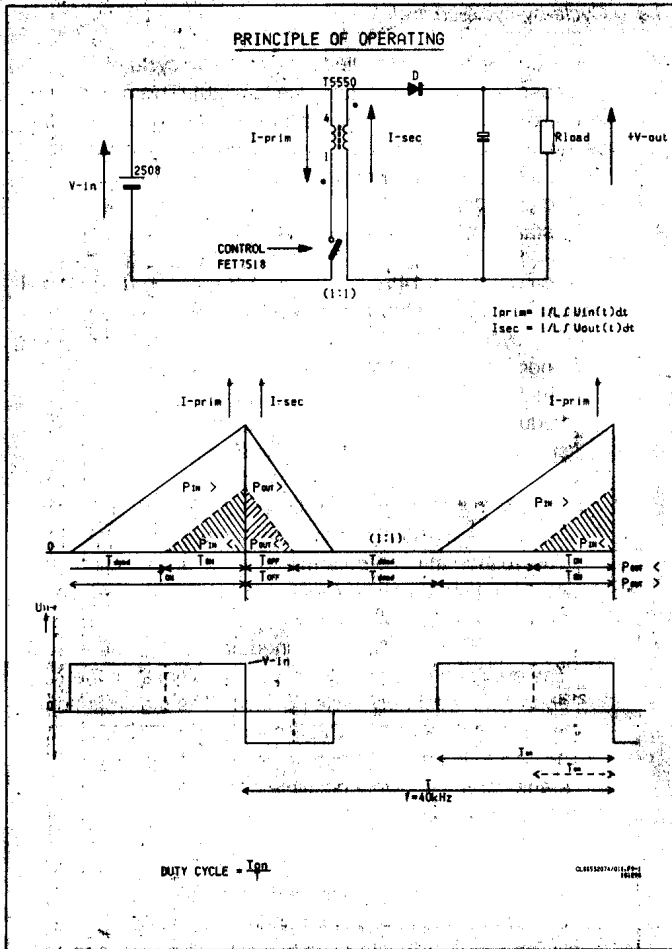


Fig. 9.1

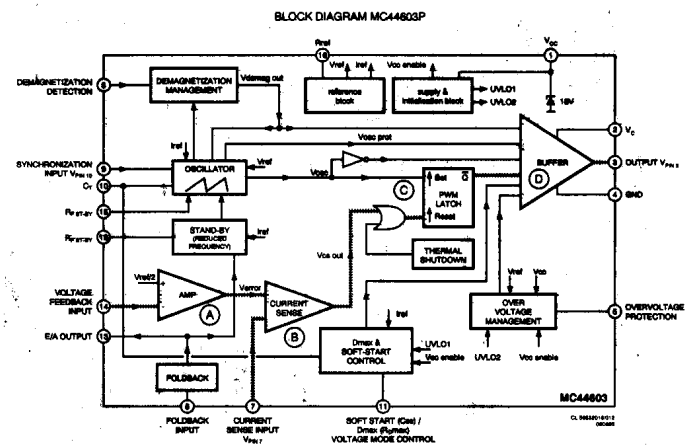


Fig. 9.2

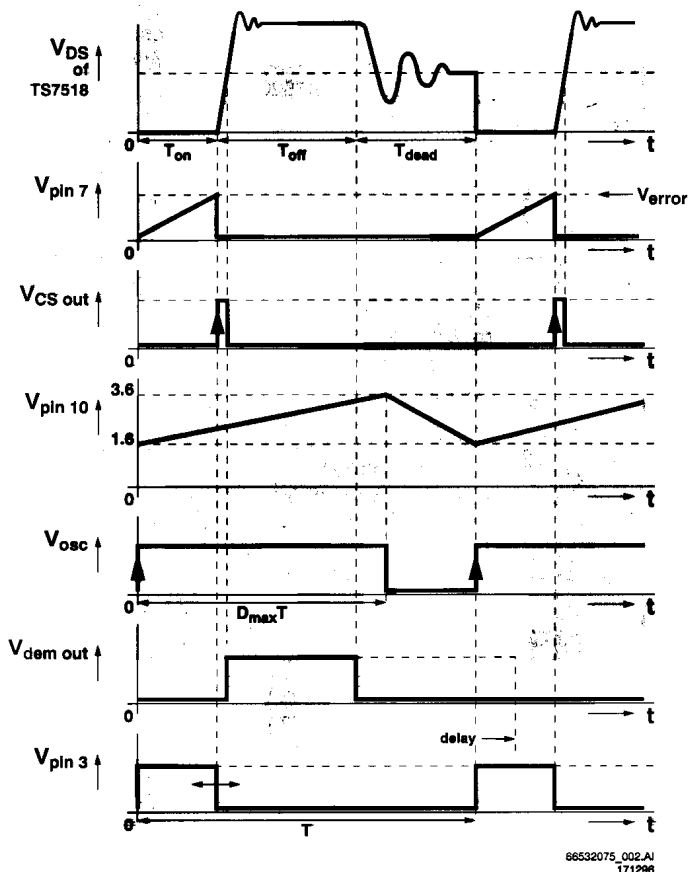


Fig. 9.3

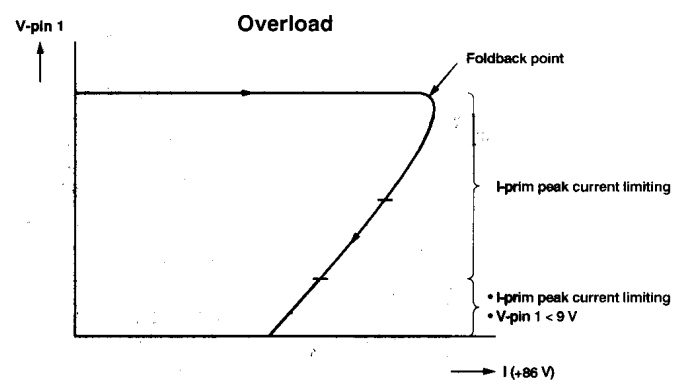


Fig. 9.4

### 9.3.5 IC7520 control (see Fig. 9.2 and Fig. 9.3)

The error amplifier (block A in Fig 9.2) compares the feedback voltage  $V_{pin14}$  with an internal reference voltage of 2V5. The output voltage  $V_{error-out}$  of this error amplifier is fed to another comparator (block B in Fig 9.2). This comparator compares the  $V_{error-out}$  and the current sense voltage  $V_{pin7}$ . As soon as the current sense voltage  $V_{pin7}$  becomes higher than the output-voltage of the error amplifier  $V_{error-out}$ , the comparator B gives a spike (the output of comparator B is the so called current sensing output-voltage  $V_{cs out}$ ).

### 9.3.6 Flip flop

Flip flop (block C in Fig 9.2) drives the output pin 3 ( $V_{pin3}$ ) via a buffer amplifier (block D). The flip flop is set by positive edge of the output of the oscillator ( $V_{osc}$ ) and reset by the spike  $V_{cs out}$ . As a result the pulse  $V_{pin3}$  becomes "high" (T-on starts) by the positive edge of  $V_{osc}$  from the internal oscillator and "low" (T-on stops) by the spike of  $V_{cs out}$  (the T-on start will be delayed in case the transformer is not yet demagnetised; see the slow-start procedure).

### 9.3.7 Stable load and increasing / decreasing load (see Fig. 9.3);

In case of a stable load, the feedback voltage  $V_{pin14}$  (and so also the maximum current sense voltage  $V_{pin7}$ ) remains the same. As a result the T-on and so the duty cycle will remain the same.

In case of an increasing load, the secondary output voltages decreases. The voltage on pin 14 would like to decrease which causes  $V_{error-out}$  to increase. As a result comparator B will give the pulse later;  $V_{pin3}$  will be "high" for a longer period (longer T-on so the duty cycle increase) and so the secondary output voltages will be increased (corrected). This will give a new balance of feedback voltage  $V_{pin14}$  and the internal 2V5 reference voltage, at a new larger duty cycle.

As a result of the longer T-on, the maximum I-prim increases, so more energy can be stored in the transformer. In this way more energy will be supplied to the load.

In case of a decreasing load, the secondary output voltages increases. The voltage on pin 14 would like to increase which causes  $V_{error-out}$  to decrease. As a result comparator B will give the pulse earlier;  $V_{pin3}$  will be "high" for a shorter period (shorter T-on so the duty cycle decrease) and so the secondary output voltages will be decreased (corrected). This will give a new balance of feedback voltage  $V_{pin14}$  and the internal 2V5 reference voltage, at a new smaller duty cycle.

As a result of the shorter T-on, the maximum I-prim decreases, so less energy can be stored in the transformer. In this way less energy will be supplied to the load.

In case the demagnetisation of the transformer is not finished, the positive edge from the oscillator, which will start a new cycle, will be overruled (via buffer block D) as being the starting point of T-on. As a result the T-on will be delayed and so the frequency of the SMPS will go down. This procedure is used during start-up.

### 9.3.8 Peak current limiting

Peak current limiting is realised by an internal clamp at  $V_{pin7}$  at 1V DC. Via this clamp the  $V_{pin7}$  can never exceed 1V DC and so the maximum value of I-prim (maximum current through FET 7518) is determined.

In case the load needs more than the maximum power, by then the I-prim is already at his maximum level so the SMPS will go in overload protection (see foldback principle explained at overload protection).

### 9.3.9 Cycle-by-cycle control

The T-on control is controlled on a cycle-by-cycle basis (because of the flip flop block C in IC7520). This means that in every cycle the T-on is determined again. By doing so the secondary voltages control, peak current limitation and all protections can be very accurate and fast.

### 9.3.10 Slow-start

As soon as  $V_{pin1} > 14V5$  DC the SMPS will start-up. This will be done by a slow-start procedure (both the frequency and the duty cycle will be built up during slow-start). The following 3 phenomena's take place during start-up:

- The frequency will slowly increase up to the nominal frequency (70 kHz for normal operation and 20 kHz for standby). This is realised via the demagnetisation function at pin 8; via this "DEMAG" function, FET 7518 will only be driven into conduction (T-on will only become "high") when T5545 is totally demagnetised.
- The voltage at pin 5 determines the foldback point. As during start-up this  $V_{pin5}$  is gradually built-up, the foldback point will also gradually increase (see foldback principle explained at overload protection).
- The duty cycle will slowly increase beginning at the absolute lowest duty cycle possible. The maximum duty cycle is determined by C2530 at pin 11 IC7520; as C2530 is uncharged at start-up, the power supply starts up at the lowest possible duty cycle.

### 9.3.11 Standby mode

In standby mode the load decreases (see description of standby on the secondary side) under a certain threshold level. The SMPS will determine this threshold level and so switch to the so called "reduced frequency mode" at 20 kHz. This minimal load threshold level is determined by R3532 at pin 12 (in the L7 the SMPS does not have a burst mode in standby, only a reduced frequency mode).  
70 kHz; In normal operation mode the internal oscillator gives 70 kHz. This frequency is controlled by C2531 at pin 10 IC7520 and by R3537 pin 16 IC7520.  
20 kHz; In standby mode the internal oscillator gives 20 kHz. This frequency is controlled by R3536 at pin 15 IC7520.

### 9.3.12 FET 7518 gate regulation

D6524 prevents pin 3 of IC7520 from becoming negative (this will destroy the IC) due to stray inductance in the gate part. The safety resistor R3525 limits the drive current to the gate of FET 7518.

### 9.3.13 Typical values for the L7 chassis

In a stable situation  $V_{pin14}$  is typical 2V5.

Mains Voltage:	110V
	220 - 240V
	150 - 276V
	90 - 276 V
Mains frequency:	50 Hz
	60 Hz
Power Consumption	
in normal mode:	14": 43 W +/- 10%
	20": 52 W +/- 10%
	21": 57 W +/- 10%
Power Consumption	
in stand-by mode:	< 10W
	< 3W option.

# Circuit description new circuitries

## 9.4 Protections

### 9.4.1 Overvoltage protection of the secondary voltages

After start-up is the supply voltage  $V_{pin1}$  taken over by positive winding 1-2, and so after start up  $V_{pin1}$  is a measuring point for the secondary output voltages. After start-up (via an internal switch) this  $V_{pin1}$  is internally tapped (voltage divided) to a voltage which can be measured at pin 6 (so  $V_{pin6}$  is also a measuring point for the secondary output voltages).

As soon as the voltage  $V_{pin6} > 2V5$ , the logic in IC7520 will shut down the output at pin 3. This 2V5 threshold at  $V_{pin6}$ , is equivalent to a  $V_{pin1}$  of 16V DC which is equivalent to a voltage at the supply voltage +VBATT of approx. 108V DC (normal operation) and 130V DC (standby). After switching "off" because of overvoltage protection, the IC starts up again (see slow-start).

→ In case an overvoltage situation is sensed at the secondary output voltages, the SMPS will go in overvoltage protection. In case the overvoltage situation remains present, the SMPS will give overvoltage protection, slow-start, overvoltage protection, slow-start, etc. → a very good audible hick-up mode.

### 9.4.2 Undervoltage protection of the secondary voltages

If the supply voltage  $V_{pin1} < 9V$  DC the output pulse at pin 3 will be shut down. As soon as  $V_{pin1} < 7V5$ , the IC7520 will be totally shut "off".  $V_{pin1}$  of 9V DC is equivalent to a voltage at +VBATT of approx. 70V DC (normal operation) and 95V DC (standby),  $V_{pin1}$  of 7V5 is equivalent to a voltage at +VBATT of approx. 55V DC (normal operation) and 65V DC (standby).

→ In case an undervoltage is sensed at the secondary output voltages, the SMPS will first switch "off" the pulse and then switch "off" the complete IC7520.

In case the IC7520 is switched "off", the SMPS will switch "off". In case the undervoltage situation remains present, the SMPS will give undervoltage protection, slow-start, undervoltage protection, slow-start, etc. → a very good audible hick-up mode.

### 9.4.3 Unload protection

In case the load goes down (e.g. the line goes down because of standby mode or some failure in the line) this is detected by IC7520 via I-prim and secondary output voltages sensing.

In case the load decreases below a certain threshold the SMPS will switch in "reduced frequency mode" of 20 kHz (this threshold is determined by the voltage level at pin 12 IC7520);

→ In case of an unload situation the set will switch to "low frequency mode" or standby mode.

Whether this unload situation of the SMPS is caused by the standby command or by a failure (e.g. in the line), can only be determined by switching on the set again which the remote control; in case of standby mode the TV will switch "on" again, in case of an unload situation the set will not switch "on".

### 9.4.4 Overload (short-circuit) protection (see Fig. 9.4)

If the secondary load becomes too high, I-prim becomes too high which is sensed by the current sense voltage  $V_{pin7}$ . This voltage  $V_{pin7}$  is not allowed to exceed 1V DC by IC7520 and so gives current limiting. As the I-prim is limited, the secondary output voltages will also drop and so supply voltage  $V_{pin1}$  will drop. As soon as  $V_{pin1} < 9V$  DC the driving pulse at pin 3 will stop.

As a result of these 2 mechanism in case of an overload the secondary voltages will drop very fast. This is called the foldback mechanism; the foldback point can be adjusted by pin 5 IC7520 (for the L7 this point is adjusted to a maximum tolerable output power of 85W at 90Vac and 165W at 276VAC).

After this foldback, the IC starts up again (see slow-start).

In case the overload situation remains present, the SMPS will give foldback again, slow-start, foldback, slow-start, etc.;

→ As a result in case of a short-circuit (or overload) the TV will be in a very good audible hick-up mode.

## 9.5 Secondary side

### 9.5.1 Output voltages

See 9.1.2 for output voltages.

### 9.5.2 Protections

No protections are available at the secondary side.

General: IC7225 (TDA836X) is a single-chip video processor with built in IF-detector, luminance-chrominance-synchronisation separator, PAL chrominance decoder, video controller, horizontal & vertical synchronisation processor en FM sound-decoder. IC7225 has 4 possible executions:

- TDA8360 is for PAL-only sets without external switch (no AV cinches)
- TDA8361 is for PAL-only sets with external switch (with AV cinches)
- TDA8362 is for PAL/SECAM multi sets with external switch (with AV cinches)
- TDA8363 is for NTSC only.

## Deflection and synchronisation (diagram A2 and A3)

### 9.6 Horizontal synchronisation IC7225-6E and the line output stage

#### 9.6.1 Synchronisation

Start up of the horizontal oscillator via the +10V gives a start-up current into pin 36; if the voltage on pin 36 exceeds 5V6 the horizontal oscillator starts running at approx. 25kHz. Only when the supply pin of IC7225 (pin 10 at IC7225-6B in diagram A7) becomes 8V the line frequency changes to 15625 Hz.

Horizontal synchronisation separator separates horizontal pulses out of CVBS and so synchronises the free-running horizontal sawtooth generator.

Horizontal oscillator sawtooth is converted into square wave voltage with variable duty cycle. This square wave on pin 37 is fed to the line output stage. The time constant of the synchronisation circuit is automatically internally determined by IC7225-6E.

## Circuit description new circuitries

Pin 38 is both SANDCASTLE output and HORIZONTAL FLYBACK input and PROTECTION input. Selection between input and output is automatically determined by the values of the current by R3456, R3462 and R3461:

- The SANDCASTLE has an output current a few mA; the amplitudes of sandcastle pulse; burst 5V3, line blanking is 3V, frame blanking 2V.
- When the input acts as a HORIZONTAL FLYBACK pulse, the input has a current of 100-300 mA. This horizontal flyback pulse compares phase of flyback pulse with phase of the horizontal oscillator. If the phase is not correct the duty cycle of horizontal oscillator will be adjusted.
- The PROTECTION signal from the frame amplifier (pin 7 IC7401 diagram A2) will be constantly "high" (see description frame amplifier) in case of no vertical deflection current. This constant "high" level will overrule the "normal" SANDCASTLE signal and so the picture will become "black".

### 9.6.2 The line output circuitry

In principal the line output stage is the same as used in the Anubis S: Pin 37 IC7225-6E drives the line output stage, TS7445 and transformer 5445 via drivers TS7440-7441. The line output stage supplies the deflection current and the following supply voltages (see also the power supply block diagram in chapter 5):

- EHT, +160, Vg2, focus and ff for the picture tube.
- +5V5 for the tuner and to create +VB for band switching.
- +9V for making the supply voltage +8V and +8VI.
- +8V and +8VI for the supply of the IC7225.
- +26V for the frame amplifier and the IC7225.

### 9.6.3 Principle working of the line output stage (see Fig 9.5)

The voltage across C2450 is constantly +95V DC. C2450 is charged by the +95V from the power supply via the primary winding 2-1 of the LOT (5445) and via R3454.

- **Second half of the scan (t1-t2):** During the second half of the scan the control voltage of TS7445 is positive, so TS7445 conducts. The horizontal deflection coil by then is switched in parallel with C2450 (constant +95V DC). As a result of this constant +95V DC a linear current is flowing through the horizontal deflection coil and TS7445. As soon as the control voltage of TS7445 becomes negative, TS7445 will not conduct any more and the second half of the scan is finished.
- **First half of the flyback (t2-t3):** During the first half of the flyback TS7445 does not conduct any more. The current which flows through the horizontal deflection coil, would like to remain flowing and so flows via C2445 bringing energy from the horizontal deflection coil to C2445. The current through the deflection coil will drop and the voltage across C2445 will rise sinusoidally.
- **Second half of the flyback (t3-t4):** During the second half of the flyback TS7445 still does not conduct. All energy which has been stored from the deflection coil into C2445 (during t2-t3) will be recovered to the deflection coil again during t3-t4. In other words, all energy in C2445 will be fed back to the horizontal deflection coil, so the voltage across C2445 drops and the current through the deflection coil will drop further (negative by now) sinusoidally.

- **First half of the scan (t4-t5):** At the end of the flyback (t4), the voltage at the cathode of the diodes D6445/D6447 parallel to TS7445 wants to become negative, so these diodes will conduct. Again the horizontal deflection coil by then is switched in parallel with C2450 (constant +95V DC). As a result of this constant +95V DC a linear current is flowing through the horizontal deflection coil and diodes D6445/D6447.

At the end of the first half of the scan the voltage at the cathodes of the diodes D6445/D6447 will become 0V, so this diodes will stop conducting. Because of that, already before the end of the first half of the scan the control voltage  $U_{BE}$  of TS7445 must be "high" again.

Horizontal flyback; The horizontal flyback pulse is brought to the correct DC level by R3456, R3462 and R3461.

D6461 prevents the pulse from becoming higher than 8V by clamping.

Horizontal S-correction to correct errors in horizontal linearity via C2450.

### 9.7 Vertical synchronisation IC7225-6E and the frame amplifier IC7401

#### 9.7.1 Synchronisation

Vertical synchronisation separator separates frame synchronisation pulses from CVBS signal and synchronises frame oscillator. The amplitude of the sawtooth on pin 43 is controlled via pin 41 (VFB vertical feedback) which looks at the vertical scan across R3410.

Pre-amplifier in IC7225-6E amplifies sawtooth (pin 43 of IC7225-6E).

#### 9.7.2 Frame amplifier

In principal the frame output stage is the same as used in the Anubis A: IC7401 (TDA3653) is used for the vertical deflection. This IC is controlled on pins 1 and 3 by the vertical control signal of IC7225-6E and a deflection current is generated on pin 5. The picture centring is set with the resistor 3409 and the picture amplitude can be set using potentiometer 3410. The vertical flyback signal is generated on pin 8 of the IC.

- During the scan the +26V supply voltage is used for the deflection current.
- During the flyback a flyback generator is used for the high  $di/dt$ . During the scan, pin 8 IC7401 is 0V and so C2403 is charged to +26V. During flyback IC7401 gives a +26V pulse on pin 8 IC7401 and so pin 6 IC7401 has a  $26+26=52V$  pulse during flyback. As a result D6403 is blocked during flyback. Since the flyback pulse at output pin 5 IC7401 is slower than at the input pin 1 IC7401 because of the self-inductance of the vertical deflection coil, a negative voltage is formed on pin 1 IC7401 during flyback. This negative voltage drives IC7401 to maximum, so the full 52V occurs on pin 5 IC7401 during flyback.
- **Protection:** In case of no deflection current, by then the flyback generator can not make +52V. As a result pin 8 will drop under 2V DC. If pin 8 drops under 2V DC the protection circuit inside IC7401 will be activated making the protection signal line on pin 7 IC7400 constant "high". This constant "high" protection will overrule the "normal" SANDCASTLE signal; the constant "high" SANDCASTLE signal will block the chrominance decoders (IC7225-6D and IC7245 in diagram A7) and so the picture will become "black".

Vertical S-correction; C2404 gives a parabolic voltage during the scan. A part of this voltage is integrated by R3418 and C2408 causing a superimposed "S-shaped" current over the deflection current which corrects the vertical linearity of the scan.  
For teletext non-interlaced mode (so 25 Hz frame) is required. For that a 25 Hz block-shaped NIL signal from the teletext decoder to the frame amplifier to ensure that odd & even frames coincide.

## Video processing (diagram A4, A7 and B1)

### 9.8 Tuning system

The tuner U1000 can be of a VST or a PLL type. In both cases the tuner is controlled by the  $\mu$ C:

- The VST tuner is controlled via V\_TUNE, AFC and the BS1 and BS2 band switching signals.
- The PLL tuner is fully I<sup>2</sup>C controlled.

### 9.9 IF demodulation IC7225-6A

IC7225-6A contains the IF amplifier and the IF detector. The IF signal is present at the output pin 11 of the tuner.

#### 9.9.1 IF band pass filter

The IF band pass characteristic is determined by the band pass of the SAW filter 1015:

- For PAL BG sets a SAW filter with 5.5 MHz bandwidth is used (33.4 to 38.9 MHz).
- For PAL I sets a SAW filter with a bandwidth of 6.0 MHz is used (32.9 to 38.9 MHz).
- For PAL BGI/SECAM BGILL' sets a SAW filter with 6.5 MHz bandwidth is used to enable BGILL' reception (33.9 to 40.4 MHz).
- For PAL BG/SECAM BGDK sets a SAW filter with a bandwidth of 6.5 MHz is used (32.4 to 38.9 MHz).
- IF-demodulator

After the band pass filter the IF signal is supplied to the IF-detector IC7225-6A pins 45 and 46. IF-demodulation is performed via the demodulation reference circuit 5260 on pins 2 and 3 IC7225-6A.

Delayed AGC control via the AGC voltage on pin 47 (AGC control is used for decreasing the amplification of the tuner-amplifiers in case the incoming signal on pin 45-46 IC7225-6A becomes too high (above the take-over level)). This take-over level can be adjusted on pin 49 by R3264.  
AFC (Automatic Frequency Control) signal on pin 44 is obtained from the reference signal of the IF-detector.

### 9.10 IF source select, luminance-chrominance separation IC7225-6B

#### 9.10.1 Sound trap

The baseband CVBS signal of pin 7 IC7225-6A (nominal amplitude of 2V<sub>pp</sub>) also contains the FM sound signal (FM intercarrier sound). This sound signal is filtered out with a ceramic filter (1206 resp. 1207) giving V-INT which is used for further video processing (IC7225 and IC7245), AV video out and teletext processing.

#### 9.10.2 Luminance-chrominance separation

Chrominance signal is filtered (-20dB) by a luminance notch filter which is internally calibrated at the subcarrier frequency (4.43 or 3.58 MHz). CVBS information is also fed to the horizontal and vertical synchronisation separator in IC7225-6E.

#### 9.10.3 CVBS source select

The V-INT signal is fed to pin 13 IC7225-6B to the source selector switch in IC7225-6B. Pin 16 is used for source select control:

- Pin 16 = 0V gives internal CVBS mode, so V-INT from pin 13 IC7225-6B
- Pin 16 = 8V gives external CVBS mode, so V-EXT from pin 15 IC7225-6B (from the video-in cinch).
- Pin 16 is DC controlled via the INT/EXT signal from buffer TS7240 which is controlled by the AV-signal of the  $\mu$ C; so AV is "high" for internal CVBS and "low" for external CVBS.

#### 9.10.4 Sharpness control

Sharpness control is realised via input pin 14 IC7225-6B (2V5-5V). Pin 14 is used as an input pin for sharpness control and an output pin for TRANS\_ID (transmission identification).

- If IC7225-6E has horizontal synchronisation (video identification), pin 14 > 0V3 and by then is input pin for sharpness control by controlling the gain of the internal luminance signal. As pin 14 > 0V3 TS7269 does not conduct and TRANS\_ID is "high" via pull-up resistor R3601 in the control part.
- If IC7225-6E has no horizontal synchronisation (no video identification), pin 14 is output pin < 0V3 so TS7269 conduct so TRANS\_ID becomes "low"

### 9.11 Chrominance decoding IC7225-6C and IC7245

PAL and NTSC chrominance decoding is inside IC7225-6C and SECAM chrominance decoding is in IC7245. PAL or NTSC processing is determined automatically by the burst demodulator inside IC7225-6C. The reference crystals for demodulation for IC7225-6C are present at pin 34 and/or pin 35 of IC7225-6C.

- PAL/NTSC mode if voltage at pin 27 < 5V5; If IC7225-6C detects PAL, the voltage at pin 27 makes no sense. If IC7225-6C detects NTSC the voltage at pin 27 is used for hue control (0-5V). For NTSC sets jumper 9246 is added.
- For Tri-Norma sets the set selects (auto or forced) one of the three different crystals for PAL M, PAL N and NTSC M at pin 34 of IC7225-6C; For Tri-Norma sets pin 26 of IC7225-6D has a double function: Saturation control (normal input pin) or Tri-Norma system select (output pin) during system search.
- PAL/NTSC/SECAM mode if voltage at pin 27 of IC7225 is 5V5; IC7225-6C searches for PAL and IC7245 searches for SECAM. Via a bi-directional communication line between pin 32 of IC7225-6C and pin 1 of IC7245, both IC's know whether a PAL/NTSC or a SECAM signal is detected. The following signals are present on the communication line:

- ⇒ 4.43 MHz signal for locking the PLL and chrominance cloche filter of IC7245.
- ⇒ SECAM or PAL/NTSC operation switching signal (DC-controlled) to do an automatic selection between the output of IC7225-6C and IC7245.  
If IC7225-6C has detected PAL or NTSC, pin 32 of IC7225-6C becomes 1V5 and the output becomes available at pin 30 and 31. If no PAL/NTSC is detected, pin 32 of IC7225-6C becomes 5V and the output will be disabled.  
If a SECAM signal is detected pin 1 of IC7245 becomes "low". This will sink current from pin 32 of IC7225-6C. In this way IC7225-6C knows that a SECAM signal is present and will disable the IC7225-6C output.

## 9.12 Video controller IC7015-6D

RGB-de-matrixing de-matrixes the -(R-Y), -(B-Y) and the Y signals to RGB signals; the sandcastle pulse coming internally from IC7225-6E synchronises the RGB de-matrixing and suppresses the RGB signals during line and frame flyback.  
Analogue controls by the  $\mu$ C for contrast (0-4V5), brightness (0-4V5) and saturation (0-4V5).  
Fast blanking and RGB-source select; Via the BL\_TXT\_OSD signal on pin 21 of IC7225-6D both the fast blanking and the RGB source select is realised via the BL\_TXT\_OSD fast blanking signal from the teletext + OSD part of the  $\mu$ C; this signal is "high" (> 1V) to switch the RGB source select switch into external mode to display teletext and OSD (via pins 22, 23 and 24 IC7225-6D).  
BCI; If the beam current increases, the BCI-signal (Beam Current Info) decreases. If the beam current is too high, the CONTRAST control signal is pulled down to reduce the contrast (pin 25 of IC7225-6D).

## 9.13 AV input cinches (diagram A6)

AUDIO-IN is an incoming audio signal from the audio-in cinch. This signal goes to source select of IC7225-6F.  
AUDIO-OUT is an outgoing audio signal from pin 1 of IC7225-6F to the audio-out cinch.  
VIDEO-IN becomes V-EXT and is the incoming CVBS-signal from the video-in cinch to the external input pin 15 IC7225-6B and the teletext processing.  
VIDEO-OUT is coming from V-INT and is an outgoing CVBS-signal taken from after the sound trap (so after the IF detector IC7225-6A) which is fed to the video-out cinch.  
The V-INT signal from the IF-detector is buffered by TS7150 before fed to the audio-out cinch.

## 9.14 CRT panel

RGB amplification by TS7300, TS7310 - TS7320, TS7330 - TS7340, TS7350 respectively  
Cut off point adjustment for adjusting the R, G and B guns to start and stop emitting at the same correct level. Via R3350, R3310 and R3330 the DC level of the collectors TS7340, 7300 and 7320 and so the DC level of the guns are adjusted.  
White D adjustment for adjusting the correct balance between R, G and B signal.

- Via R3340 and R3300 the amplitude of B and R signal can be adjusted to the amplitude of G
- Via TS7360 the R3340 and R3300 adjustment is de-coupled from influencing the G-amplification; the base DC-voltage of the RGB-amplifiers is equal to the black level of the RGB signals

Picture tube flash protection:

- Spark gaps in the PWB of the picture tube panel
- Resistors in series with the RGB electrodes 3355, 3215 and 3335 limiting the current through the guns

- Diodes 6354, 6314 and 6334 conduct at flash-over and so do not allow a higher voltage at the guns as approx. 160V  
Peak beam current limiter; If the beam current is too high, the current through resp. R3352, 3312 and 3332 is high. The diodes 6350, 6310 and 6330 conduct and so TS7350, 7310 and 7330 can not supply more current to the guns and so the beam current is limited.

## Audio processing (diagram A6 and A8)

### 9.15 FM and AM demodulation

Two sound paths can be determined:

- For BG, I, DK, M and N systems FM modulated inter-carrier sound (sound extracted from baseband CVBS from IF detector)
- For LL' systems AM modulated quasi-split sound (sound extracted directly from the tuner).

#### 9.15.1 FM demodulation

For FM modulated sound the sound signal is filtered through filter 1101 or 1102 from the baseband CVBS signal.

Input characteristic; By the switching signal CHROMA\_1/I/BG/L/DK transistor 7102 can be switched on/off.

- In case CHROMA\_1/I/BG/L/DK is "low", TS7102 does not conduct and filter L1102 is switched in parallel to L1101.
- In case CHROMA\_1/I/BG/L/DK is "high", L1102 is not in parallel with L1101 any more. The frequency of the filters is mentioned on it.

FM-mono sound demodulation takes place in IC7225-6F. No adjustment is required for BG or I demodulation as automatic PLL tuning (4.2 to 6.8 MHz) is used. Pin 1 of IC7225-6F is used as:

- input for defining the sound frequency characteristic by de-emphasis C2101
- output for feeding the FM demodulated sound.  
Source select between FM sound or AUDIO IN sound (pin 6 IC7225-6F) is done via pin 16 IC7225-6B (diagram A7).

#### 9.15.2 AM demodulation

AM-sound is for the moment not applicable. If in the future AM-sound becomes available this will be described.

### 9.16 Audio control and amplification

Bass and treble are directly controlled by the micro-controller. The bass signal is "low" for switching the bass amplification on. The treble signal is "low" for switching the treble amplification on. If bass amplification is "off", 7124 is short-circuiting resistor 3124. If treble amplification is "off" resistor 3117 and capacitor 2117 are short-circuited by 7116.  
Audio amplification is realised via the sound-amplifier 7120 or 7121 (depending on the version). The only difference is the output power.

## Control and teletext (diagram A5):

### 9.17 Teletext

In the L7 two microprocessors can be used; one with and one without teletext.

- In case of TXT, this teletext function is integrated together with the control part in one and the same  $\mu$ C. This  $\mu$ C is drawn in the diagrams with the outern pin numbering.
- In case of no TXT another  $\mu$ C is used with less pins. This  $\mu$ C is drawn in the diagrams with the internal pin numbering.

In the description below, the pin numbers mentioned are the numbers mentioned outside the housing of IC7601, so for the  $\mu$ C with integrated TXT functionality. In case of the  $\mu$ C with integrated teletext function, the CVBS-signal is fed to pin 23 or 24 depending on the fact if it is the internal or external CVBS-signal (V\_INT or V\_EXT). In this way teletext can be used both on the internal or the external signal. The TXT and OSD-information is combined at pins 32-33-34.

## 9.18 Control

Following description explains the functionality of the  $\mu$ C pins anti-clockwise for the outern pinning numbers.

- Control-voltage outputs (pin 1-7 and pin 9-10); These pins are PWM (Pulse Width Modulated) output pins used for volume, contrast, saturation, hue, brightness, sharpness, bass and treble and tuning control (only for VST).
  - ⇒ The V-TUNE varies between 0-30V and is derived from the +95V supply from the power supply.
  - ⇒ The saturation pin 4 has two functions; output pin for saturation control and input pin for auto system search in case of Bi- and tri-norma sets (-/77 sets).
  - ⇒ Bass and treble functionality is only used in case of sets with the "smart sound" feature.
- AV (pin 8); Output switching signal "high" for internal CVBS-mode and "low" for external mode (AV-mode, so cinch mode).
- AFC (pin 11); Input pin for AFC-control.
- AV\_MUTE (pin 12); Output switching signal used for muting the audio output cinch. This signal is "high" in case of mute.
- Functional switch (pin 15); For USA ,sets do not have a mains switch but a functional switch. If pin 15 is connected to ground by means of 1064, the set is switched to stand-by.
- Protection (pin 16); This pin is an input pin for protections. If this pin is connected to ground, the set is switched in protection. By this protection the voltages +9V and HEW are monitored to check if they become to high. If the +9V drops, this is monitored by the circuit around 7608. The emitter becomes "low" (0V7 lower than the base voltage) if the +9V drops. This will force pin 16 of the  $\mu$ C "low" and will switch the set in protection.
- BS1 and BS2 (pin 17-18); Switching signals used for band switching of a VST tuner.

	BS1	BS2
VHF1	0	1
VHF2	1	0
UHF	1	1

- STANDBY (pin 19); Output pin "high" for normal operation and "low" for standby.
- LED-drive (pin 20); Signal to drive the LED
  - ⇒ In standby, the LED lights continuously by pulling pin 20 "low"
  - ⇒ In normal operation the LED does not light by not pulling pin 20 "low"
  - ⇒ During RC5 reception pin 20 is pulled "low" time by time, resulting in a pulsing LED
- Ground (pin 21); Ground of the power-supply.
- Test pin (pin 22); Used for test purposes in the factory
- CVBS-inputs (pin 23-24); These pins are used as input for teletext-sources. Pin 24 is used as input for the external CVBS-signal (VIDEO-IN input cinch) and pin 23 for the internal CVBS-signal of the set.
- NIL (pin 27); Signal to generate a DC-current through the deflection coil to create a non interlaced mode during TXT-mode.

- TXT/OSD-signals (pin 32-33-34); These output pins are used to create TXT and OSD information in different colours.
- BL-TXT-OSD (pin 35); Output signal (BL\_TXT\_OSD) used to indicate the video controller that there is OSD or Teletext information. So this signal blanks the video information.
- SANDCASTLE (pin 36); Pin to inform the  $\mu$ C that horizontal flyback takes place. This information is needed to place the TXT and OSD correctly on the picture.
- VFL (pin 37); This pin is used to tell the  $\mu$ C that vertical flyback takes place. This information is needed to place the TXT and OSD correctly on the picture.
- OSD-generator (pin 38-39-40); The components connected these pins determine the frequency of the OSD-generator. This is approx. 8 MHz.
  - ⇒ In a non TXT set, the OSD generator is formed by C2680, C2681, L5680 and L5681 (4682 and 4683 are not mounted).
  - ⇒ In a TXT set, C2680, C2681 and L5680 are not present but 4682 and 4683 are mounted.
- 12 MHz oscillator (pin 41-42); The frequency of the oscillator of the  $\mu$ C is determined by this crystal 5600.
- POR (pin 43); At switching on the set with the mains switch the signal at pin 43 becomes "high" and holds the  $\mu$ C. The  $\mu$ C waits until the signal at pin 43 becomes "low". In this way the  $\mu$ C knows that the supply-voltage is high enough to be able to perform well.
- TXT / no TXT (pin 44); In case jumper 4602 is present, the software "knows" as a no-TXT set (PCF84C44). In case jumper 4602 is not present, the software "knows" as a TXT set (SAA5290).
- IR-input (pin 45); Input for the remote-control commands
- Video system selections (pin 46-47-48); These three outputs can be used in different ways depending on the region where the set is produced for:
  - ⇒ For Asian Pacific sets the CHROMA1\_I/BG/L/DK signal is used for sound crystal selection in the FM sound demodulation part. In case I/BG/L/DK signal is "low" L1102 is switched in parallel to L1101.
  - ⇒ For Latin America a so called Bi-Norma (PAL-M and NTSC-M) or Tri-Norma (PAL M/N and NTSC M) is configured by using the CHROMA\_0, CHROMA\_1 and CHROMA\_2 switching signals. For these Bi- and Tri-Norma sets the SATURATION output pin 4 is also used as an input pin for the Tri-Norma automatic system selection.

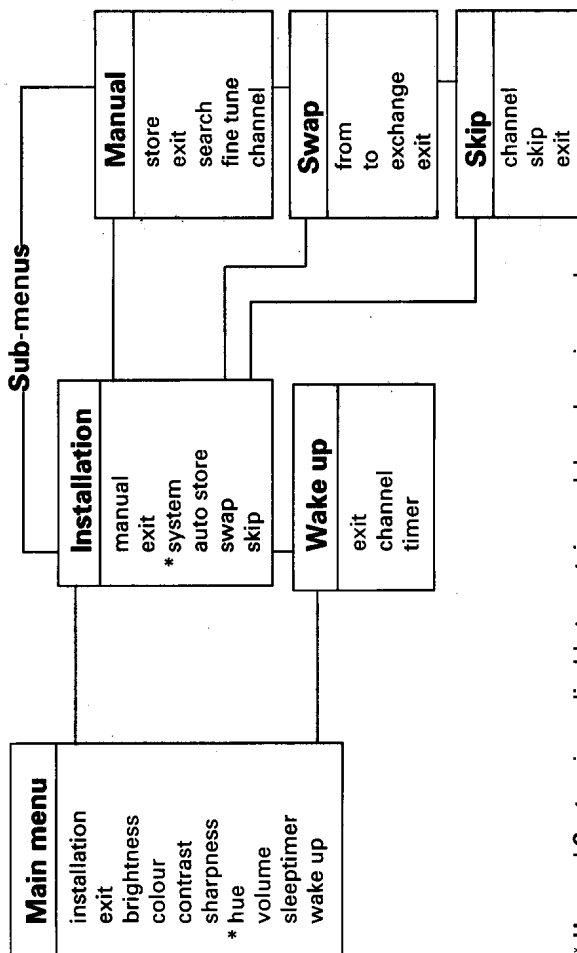
	CHROMA_0	CHROMA_1	CHROMA_2
PAL M	0	1	
PAL N	1	0	
NTSC M	1	1	

- I<sup>2</sup>C-Bus (pin 49-50); This bus is used to communicate with all used I<sup>2</sup>C devices.
  - ⇒ Non Volatile Memory (EEPROM) in which the settings are stored. In case pin 1 of this NVM is shorted while switching on the set with the mains switch, the SDAM (Service Default Alignment Mode); see chapter 6.
  - ⇒ In case of a PLL tuner, the I<sup>2</sup>C-Bus is used via the copper tracks of BS1 and BS2 (these copper tracks are used for band switching in a VST set).
- VIDEO\_ID (video identification; pin 51); Pin 51 is "high" in case a video signal is detected and "low" in case no video signal is detected. This signal is coming from pin 14 IC7225-6B.
- Supply voltage (pin 52); If this voltage is present and the Power On Reset (POR) signal at pin 43 is "low" the  $\mu$ C will start.



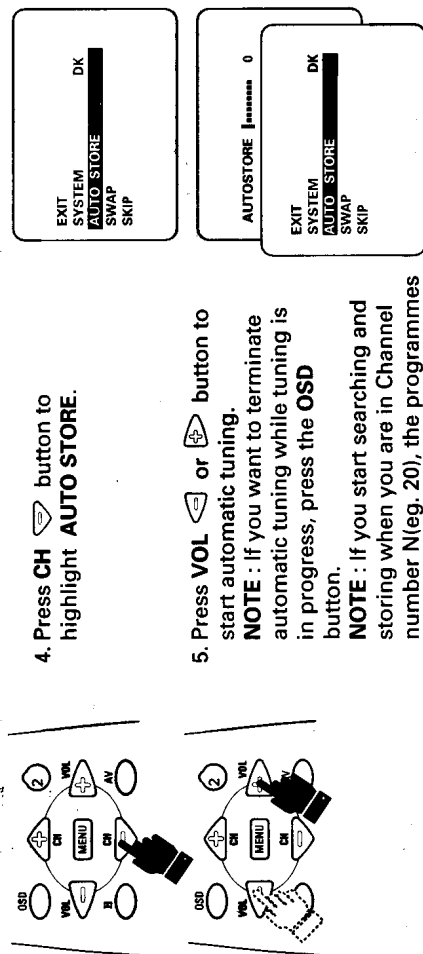
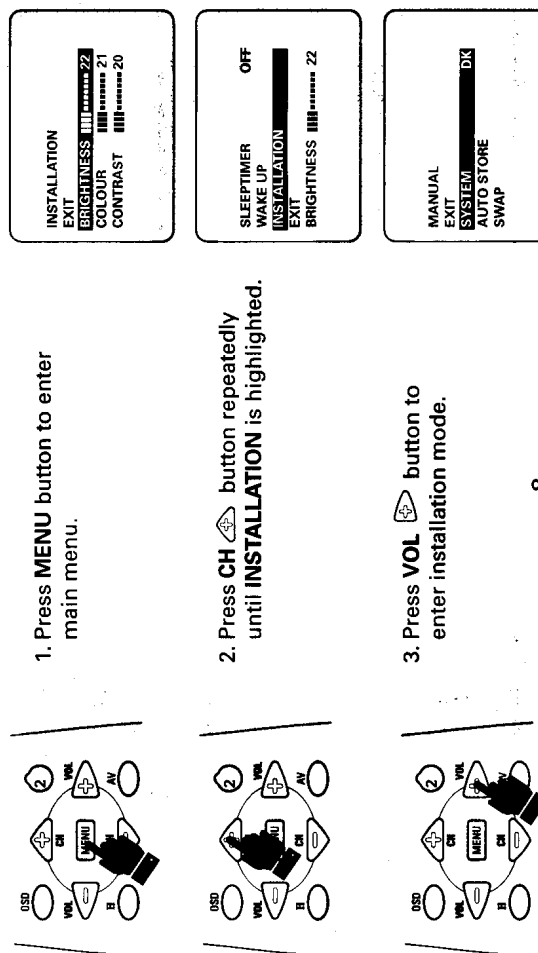
Operating instructions generally explain the operation of the TV set using the buttons on the remote control handset unless otherwise stated.

## Overview of main menu and sub-menus



\* Hue and System is applicable to certain models and versions only.

## How to start Automatic Installation (Auto Store)



**NOTE** : If you start searching and storing when you are in Channel number N (eg. 20), the programmes found will start storing from Channel N + 1 (eg. 21) onwards.

**NOTE** : Searching and storing stops at the last channel (Channel 79).

6. Once automatic tuning is completed, press **OSD** button to exit from menu.

\* To get back to main menu, you need to select "EXIT" on the sub-menu and press **VOL** or button to exit. If necessary, repeat the above procedure until main menu appears.

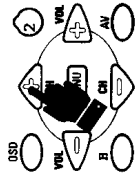
## NOTE

- (Applicable to certain models/versions only)
- If you hear any "noisy" sound on any channel after automatic tuning is completed, repeat **step 1** to **3**.
- Press **VOL** or button to select **PAL I** or **PAL DK**.
- Press **OSD** button to exit from menu.

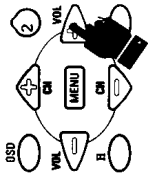
## How to start Manual Installation

You can also do installation manually by the **SEARCH** method. Manual installation allows you to select your preferred channel number for every available station

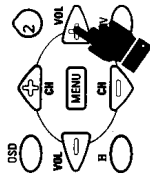
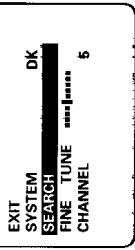




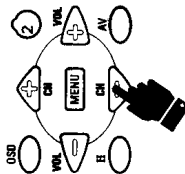
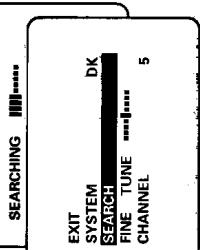
- 4 Press **CH** button repeatedly until **MANUAL** is highlighted.



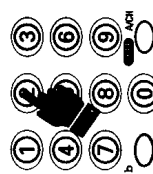
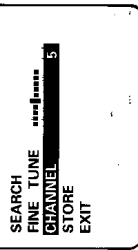
5. Press **VOL** button to enter manual mode.



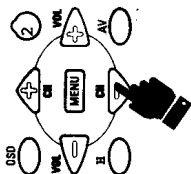
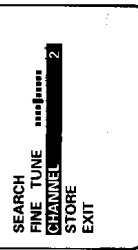
6. Press **VOL** button to activate **SEARCHING** mode. Searching stops once a station is available. If you decide to store the available station, proceed to the next step. However, if you decide to continue searching for another station, press **VOL** button again until another station is found.



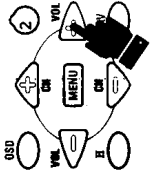
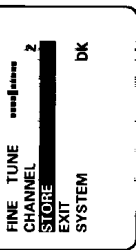
7. Press **CH** button repeatedly until **CHANNEL** is highlighted.



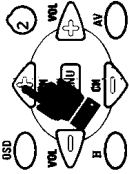
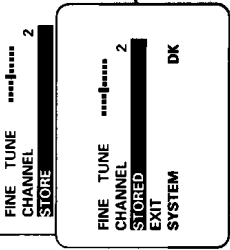
8. Key In desired channel number by the **DIGIT (0 - 9)** button.



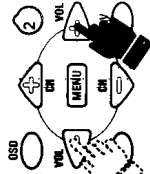
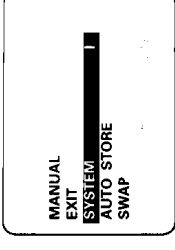
9. Press **CH** button to highlight **STORE**.



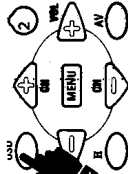
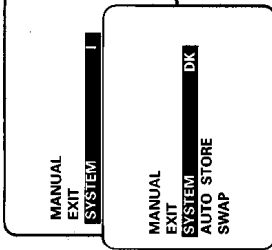
10. Press **VOL** button to store the channel.



11. If you hear any "noisy" sound after manual installation is completed, press **CH** button to highlight **SYSTEM**. (Applicable to certain models and versions only).



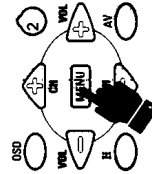
12. Press **VOL** or button to select **PAL I** or **PAL DK**. (Applicable to certain models and versions only).



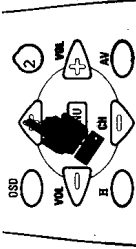
13. Press **OSD** button to exit from menu.

### How to Swap Channels

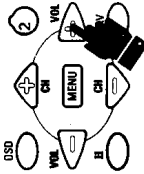
This feature allows you to change the channel number to your choice for a particular TV station.



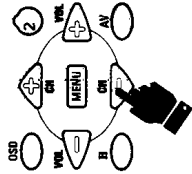
1. Enter main menu.



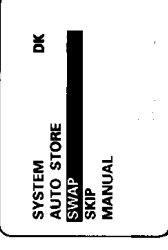
2. Highlight **INSTALLATION**.



3. Enter installation mode.



4. Press **CH** button repeatedly until **SWAP** is highlighted.



# 11. List of abbreviations (incl. all signal names)

+160V	+16V supply voltage from the LOT to the picture tube panel
+95V	+95V supply voltage from the SOPS to the line output stage and the tuning circuit
+26V	+26V supply voltage from the LOT to the frame amplifier IC7401
+10V/14V	+xxV supply voltage from the SOPS to supply the audio amplifier
+10V	+10V supply voltage from the SOPS to the line drive stage (A3)
+9V	+9V supply voltage from the LOT to the relais of the degaussing coil and to the supply voltages +8V and +8VI
+8V/+8VI	+8V supply voltage from the LOT to supply IC7225
+5V5	+5V5 supply voltage from the LOT for the tuner and to create VB for bandswitching
+5V	+5V supply voltage from the SOPS to supply the control part
$\mu$ C	Microcomputer
AFC	Automatic Frequency Control
AGC	Automatic Gain Control
AUDIO_IN	AUDIO-IN signal from audio cinch; this signal is fed to IC7015-6F for source select
AV	Switching signal from the $\mu$ C to select between internal and external video/audio
AV-MUTE	Signal to mute the sound on the Audio-out cinch
AQUA	Aquadag on the rear side of the picture tube to pin 8 of the LOT
AUDIO_OUT	Outgoing audio signal from pin 1 of IC7225-F to audio_out cinch
B_TXT_OSD	Blue input signal from the $\mu$ C to the video controller IC7015-6D
BS1	Switching signal from $\mu$ C for band switching to tuner 1000
BS2	Switching signal from $\mu$ C for band switching to tuner 1000
BCI	Beam Current Info; If beam current increases the BCI signal decreases. BCI is used for contrast reduction (if beam current is too high)
BL-TXT-OSD	Fast blanking signal to IC7225-6D to display OSD and TXT
BRIGHTNESS	Control signal (from $\mu$ C, but on DC level via RC network) for brightness control of the video controller IC7015-6D (0-5V)
CHROMA	Chrominance part of the video signal
CHROMA-0_L/L'/I	Signal to select the correct system in case of trinorma
CHROMA-1_I/BG/L/DK	Signal from the $\mu$ C to select the correct sound x-tal. In case of trinorma to select the correct system
CHROMA-2/STATUS	Signal to select the correct system in case of trinorma
CONTRAST	Control signal (from $\mu$ C, but on DC level via RC network) for contrast control of the video controller IC7015-6D
CVBS	Colour Video Blanking Synchronisation
V-EXT	Incoming CVBS signal from cinch video_in to the external input pin 15 IC7015-6B
V-INT	Outgoing CVBS signal from sound trap on pin 7 IC7015-6A (IF detector) to the video_out cinch
EEPROM	Electrical Erasable Programmable Read Only Memory
ESD	Electrical Static Discharge
ff	Filament (heater voltage) from LOT to the picture tube
FM	FM demodulated sound from the FM-demodulator IC7015-6F to smart sound
G-TXT-OSD	Fast blanking signal to IC7225-6D to display OSD and TXT
HUE	Signal from the $\mu$ C to control the hue of the video signal
HEW	X-ray detection. If this signal is too high, X-ray could occur so the set is switched in protection
HOR. FLYBACK	Horizontal flyback pulse (15625 Hz) used for locking the horizontal oscillator in IC7015-6E
I <sup>2</sup> C	Digital control bus of the microcomputer
VIDEO-ID	Status signal from IC7015-6B; "low" for no CVBS signal (horizontal sync not present), "high" in case CVBS signal is present (horizontal sync present) from the IF-detector IC7015-6B to the $\mu$ C
IF	Intermediate frequency signal from the tuner
NIL	Non Inter Lace; 25 Hz block-shaped signal from teletext to the frame amplifier for coinciding the odd & even frames
POR	Power On Reset; ensures the $\mu$ C starts up its software only if the power supply of the $\mu$ C itself is high enough
PP	Personal Preference
PROT	Protection signal from frame IC7401; in case the vertical flyback generator in IC7401 is not activated, the voltage on pin 8 IC7401 becomes < 2V. By then the protection circuit in IC7401 will make pin 7 "high" overriding the HOR FLYBACK and SANDCASTLE. The constant "high" sandcastle will cause the picture to become "black"
R_TXT_OSD	Fast blanking signal to IC7225-6D to display OSD and TXT
RAM	Random Access Memory
ROM	Read Only Memory
SANDCASTLE	Sandcastle signal from IC7015-6F to delay line IC7255 and SECAM chrominance decoder IC7245
SATURATION	Control signal (from $\mu$ C, but on DC level via RC network) for saturation control of the video controller IC7015-6D (0-2V5)
SAW	Surface Acoustic Wave; high precision band pass filter
SCL	Clock line of the I <sup>2</sup> C-bus
SDA	Data line of the I <sup>2</sup> C-bus
SAM	Service Alignment Mode; Service mode for doing alignments.
SDM	Service Default Mode; predefined mode for faultfinding (see chapter 8)
SDAM	Service Default Alignment Mode; Combined mode of SAM and SDM.
SHARPNESS CONTROL	Control signal on DC level (0-5V) from $\mu$ C to IF-detector IC7015-6B) for sharpness control

# List of abbreviations (incl. all signal names)

Chassis L7.1A

33

SMART SOUND	Bass and treble control before the sound amplifier.
STANDBY	Switching signal from $\mu\text{C}$ ; "low" for standby (power supply will be switched to stand-by mode), "high" for normal operation
INT/EXT	Switching signal derived from the AV-signal for internal or external audio + video switching ("low" for internal and "high" for external)
VT	Tuning voltage from which the signal TUNING VOLTAGE is derived to tune the tuner
VERT DRIVE	Vertical drive signal from IC7225-6E to frame amplifier IC7401
VFB	50 Hz vertical flyback pulse used for locking the vertical oscillator in IC7225-6E
VFL	50 Hz vertical flyback pulse used to inform the $\mu\text{C}$ that flyback takes place. This is important for OSD and TXT.
Vg2	Voltage on grid 2 of the picture tube
VOLUME	Control signal (from $\mu\text{C}$ , but on DC level via RC network) for volume control of sound processing in IC7225-6F
Y	Luminance part of the video signal

## Main carrier [A]

## Various

▲	4822 492 70788	SPRING
▲	4822 265 20689	CONN. 2-P MALE
▲	4822 492 70289	SPRING
▲	4822 265 20439	CONNECTOR 2-P
▲	4822 276 13603	SWITCH, MAINS
▲	4822 256 92053	PLASTIC HOLDER
▲	4822 265 20723	CONNECTOR 2-P
▲	4822 256 10336	LED HOLDER
▲	4822 157 11166	EMI FILT. 40MHz
▲	4822 257 10538	CONN. 3-P MALE

▲	4822 267 31014	PHONE CONN.
▲	4822 267 10549	CONN. 4-P FEM
▲	4822 265 10481	CINCH CONN 2-P
▲	4822 441 11878	CINCH HOUSING
▲	4822 276 13603	MAIN SWITCH
▲	4822 256 92053	FUSE HOLDER
▲	4822 157 11166	EMI FILT. 40MHz
▲	4822 267 10538	CONN. 3-P MALE
1000	4822 210 10737	TUNER UV1355/I
1015	4822 242 72197	FILTER 38MHz

1015	4822 242 73792	FILTER 45MHz
1060	4822 276 13775	SWITCH
1061	4822 276 13775	SWITCH
1062	4822 276 13775	SWITCH
1063	4822 276 13775	SWITCH
1101▲	4822 242 10316	FILTER 6.5MHz
1102	4822 242 10314	FILTER 5.5MHz
1102	4822 242 10362	FILTER 6.0MHz
1102	4822 242 10363	FILTER 4.5MHz
1206	4822 242 81572	FILTER 6.0MHz

1206	4822 242 81712	FILTER 5.5MHz
1206	4822 242 81978	FILTER 4.5MHz
1207	4822 242 81301	FILTER 6.5MHz
1275	4822 242 10356	X-TAL 4.433MHz
1277	4822 242 10355	X-TAL 3.579MHz
1449▲	4822 071 54001	FUSE 400mA
1500	4822 070 34002	FUSE 4A
1571▲	4822 071 51602	FUSE 1.6A
1572▲	4822 071 51351	FUSE 315mA
1670	4822 218 11573	IR RECEIVER

1681	4822 242 10694	X-TAL 12MHz
1681	5322 242 73686	FILTER 12MHz



2008	4822 126 13296	100nF 10% 16V
2008▲	5322 122 34123	1nF 10% 50V
2010	4822 124 11582	2200µF 20% 16V
2011	5322 122 32452	47pF 5% 63V
2013	5322 122 32452	47pF 5% 63V
2016▲	4822 124 40433	47µF 20% 25V
2101▲	5322 126 10223	4.7nF 10% 63V
2101	5322 126 10465	3.9nF 10% 63V
2102	4822 121 43897	1nF 5% 400V
2102	4822 126 13498	82pF 5% 50V

2102	4822 126 13644	47pF 5% 63V
2103	4822 126 13061	220nF 20% 25V
2104	4822 124 40248	10µF 20% 63V
2105	4822 124 81108	0.47µF 20% 50V
2106▲	4822 122 33342	33nF 10% 63V
2107▲	5322 126 10223	4.7nF 10% 63V
2108	4822 122 33515	82pF 5% 63V
2108	4822 126 13693	56pF 1% 63V
2108	5322 122 32452	47pF 5% 63V
2109	4822 124 41576	2.2µF 20% 50V

2110▲	4822 126 13838	100nF 20% 50V
2111	4822 124 81028	220µF 20% 25V
2116	4822 121 51379	82nF 5% 63V
2117▲	5322 122 32654	22nF 10% 63V
2120	4822 121 42868	220nF 5% 50V
2121	4822 126 13061	220nF 20% 25V
2122	5322 126 10511	1nF 5% 50V
2124	5322 121 42386	100nF 5% 63V
2130	4822 124 11566	47µF 20% 50V
2135	4822 124 81033	100µF 20% 50V

2136	4822 124 81033	100µF 20% 50V
2138▲	4822 121 43823	470nF 5% 50V
2150	4822 124 81022	1µF 20% 50V
2152	4822 124 41576	2.2µF 20% 50V
2190	4822 126 13512	330pF 10% 50V
2208	4822 126 13751	47nF 10% 63V
2212	5322 121 42386	100nF 5% 63V
2213	4822 126 13561	220nF 10% 16V
2221▲	4822 126 13838	100nF 20% 50V
2222	4822 124 41576	2.2µF 20% 50V

2224	4822 124 41584	100µF 20% 10V
2228	4822 126 13296	100nF 10% 16V
2231▲	5322 122 32654	22nF 10% 63V

2243▲	5322 122 32654	22nF 10% 63V
2245▲	4822 126 13838	100nF 20% 50V
2246	4822 126 13628	220N 20% 50V
2248▲	5322 122 34123	1nF 10% 50V
2249▲	5322 122 34123	1nF 10% 50V
2251▲	5322 122 32654	22nF 10% 63V
2254	4822 124 81164	22U 20% 25V

2256▲	4822 126 13838	100nF 20% 50V
2257▲	4822 126 13838	100nF 20% 50V
2260	4822 126 13689	18pF 1% 63V
2260	5322 122 33869	15pF 5% 63V
2261▲	5322 122 32654	22nF 10% 63V
2264▲	5322 122 32654	22nF 10% 63V
2265	4822 124 41576	2.2µF 20% 50V
2265	4822 124 81108	0.47µF 20% 50V
2270	4822 126 13296	100nF 10% 16V
2271	4822 126 13296	100nF 10% 16V

2272	5322 122 33446	3.3nF 10% 63V
2273	4822 126 13296	100nF 10% 16V
2275	5322 122 33869	15pF 5% 63V
2277	5322 122 33869	15pF 5% 63V
2280	4822 126 13751	47nF 10% 63V
2284	4822 126 13751	47nF 10% 63V
2285	4822 126 13751	47nF 10% 63V
2290▲	4822 126 13838	100nF 20% 50V
2291	4822 124 40849	330µF 20% 16V
2299	4822 122 40606	22nF 20% 50V

2301	5322 121 42386	100nF 5% 63V
2304	4822 126 10334	470pF 10% 50V
2324	4822 122 33528	390pF 5% 50V
2324	5322 122 32336	560pF 10% 100V
2344	4822 126 10334	470pF 10% 50V
2370	4822 121 41689	100nF 10%
2373	4822 121 41926	33nF 5% 630V
2400	4822 122 33127	2.2nF 10% 63V
2401	4822 122 32646	5.6nF 10% 50V
2402	4822 122 33528	390pF 5% 50V

2403	4822 124 41596	22µF 20% 50V
2403	4822 124 81033	100µF 20% 50V
2404	4822 124 40248	10µF 20% 63V
2404	4822 124 41596	22µF 20% 50V
2406	4822 121 43901	4.7nF 5% 50V
2407	4822 121 51399	47nF 10% 50V
2407	5322 121 42386	100nF 5% 63V
2408	4822 124 11582	2200µF 20% 16V
2408	4822 124 81039	3300µF 20% 25V
2421	4822 122 32627	2.7nF 10% 50V

2422	4822 124 81022	1µF 20% 50V
2423▲	5322 126 10223	4.7nF 10% 63V
2424▲	4822 126 13838	100nF 20% 50V
2425	4822 124 81164	22U 20% 25V
2426	5322 121 42386	100nF 5% 63V
2427▲	5322 126 10223	4.7nF 10% 63V
2440	4822 121 43925	2.2nF 5% 50V
2442	4822 126 13628	220N 20% 50V
2443	4822 124 40198	470µF 20% 16V
2444	4822 121 51319	1µF 10% 63V

2445▲	4822 121 70618	12nF 5% 1600V
2445	4822 121 70649	9.1nF 5% 1.6KV
2448	4822 121 43368	47µF 20% 160V
2450	4822 121 10506	560N 5% 250V
2450	4822 121 10507	470N 5% 250V
2451	4822 121 51319	1µF 10% 63V
2452	4822 124 81165	470U 20% 10V
2453	4822 124 11771	1000µF 20% 35V
2460	4822 121 43245	68nF 10% 100V
2460	4822 121 43378	82nF 10% 100V

2461	4822 126 13645	27pF 5% 50V
2462	4822 126 11824	100pF 10% 1KV
2470▲	4822 124 11508	22µF 20% 250V
2471	4822 121 41856	22nF 5% 250V
2500▲	4822 126 13589	470N 20% 275V
2501	4822 121 70141	33nF 5% 400V
2502	4822 126 12793	2.2nF 10% 2KV
2504	4822 126 12793	2.2nF 10% 2KV
2505	4822 126 12793	2.2nF 10% 2KV
2508	4822 124 11907	100µF 20% 400V

2508	4822 124 41748	220µF 20% 400V
2509	4822 122 50116	470pF 10% 1KV
2510▲	4822 122 50116	470pF 10% 1KV
2517▲	5322 122 34123	1nF 10% 50V
2518▲	4822 126 12426	330pF 10% 1KV
2520	4822 122 33515	82pF 5% 63V
2521	4822 122 33127	2.2nF 10% 63V
2522	4822 122 33127	2.2nF 10% 63V
2529▲	4822 126 13838	100nF 20% 50V
2530	4822 124 81022	1µF 20% 50V

2531	4822 121 10646	560pF 1% 400V
2533	5322 122 31863	330pF 5% 50V
2534	5322 126 10511	1nF 5% 50V
2537	5322 121 42386	100nF 5% 63V
2540	4822 124 81029	100µF 20% 25V
2541	4822 121 43872	3.3nF 5% 50V
2545▲	4822 126 14049	1.5nF 20% 250V

2550▲	4822 126 12426	330pF 10% 1KV
2551	4822 124 42336	47µF 20% 160V
2552	4822 126 13597	330pF 10% 500V
2561	4822 124 40198	470µF 20% 16V
2563	4822 124 41596	22µF 20% 50V
2571	4822 124 11908	2200µF 20% 25V
2572	5322 122 32531	100pF 5% 50V
2602▲	4822 124 40433	47µF 20% 25V
2610	4822 126 13628	220N 20% 50V
2611	4822 124 40248	10µF 20% 63V
2615	4822 126 13628	220N 20% 50V
2620	5322 126 10184	680P 5% 50V
2621	4822 122 33515	82pF 5% 63V

2622	4822 122 33515	82pF 5% 63V
2623▲	5322 122 32654	22nF 10% 63V
2630	4822 124 40248	10µF 20% 63V
2631	4822 124 40248	10µF 20% 63V
2632	4822 124 40248	10µF 20% 63V
2633	4822 124 40248	10µF 20% 63V
2634	4822 124 81022	1µF 20% 50V
2635▲	4822 124 40433	47µF 20% 25V
2639▲	4822 126 13838	100nF 20% 50V
2650	5322 126 10184	680P 5% 50V

2652	5322 126 10184	680P 5% 50V
2653	5322 122 34098	10nF 10% 63V
2662	5322 126 10184	680P 5% 50V
2664▲	4822 126 13838	100nF 20% 50V
2665▲	4822 126 13838	100nF 20% 50V
2666▲	4822 126 13838	100nF 20% 50V
2670	5322 122 32531	100pF 5% 50V
2671	4822 124 81029	100µF 20% 25V
2674	5322 122 32531	100pF 5% 50V
2680	4822 126 13689	18pF 1% 63V

2681	4822 126 13689	18pF 1% 63V
2682	5322 122 31946	27pF 5% 63V
2683	5322 122 31946	27pF 5% 63V
2687	4822 122 33575	220pF 5% 50V
2690	4822 122 33575	220pF 5% 50V
2691	5322 122 32531	100pF 5% 50V
2692	5322 122 32531	100pF 5% 50V
2693	5322 122 32531	100pF 5% 50V
2694▲	4822 126 13689	100nF 20% 50V

3000	4822 050 11002	1k 1% 0.4W
3004	4822 050 11002	1K 1% 0.4W
3005	4822 050 11002	1K 1% 0.4W
3006	4822 050 11002	1k 1% 0.4W
3007▲	4822 051 20102	1k 1% 0.4W
3008	4822 050 11002	1k 1% 0.4W
3009	4822 050 11002	1K 1% 0.4W
3010▲	4822 052 10478	4k7 5% 0.33W
3011	4822 051 20822	8k2 1% 0.4W
3101	4822 051 20394	390k 5% 0.

3470▲ 4822 052 11278 4Ω 5% 0.5W  
3471 4822 117 12651 22Ω 5% 2W  
3480 4822 117 12648 100Ω 5% 2W  
3499▲ 4822 052 10108 1Ω 5% 0.33W  
3500▲ 4822 117 12164 430V - 710V

3501 4822 117 12181 470Ω 20% 0.5W  
3503 4822 116 40204 30Ω 30%  
3504▲ 4822 116 40277 PTC 9Ω S 100R  
3506 4822 116 82776 2Ω 2  
3507 4822 117 12654 100Ω 5% 5W  
3510 4822 117 12647 33k 5% 3W  
3512 4822 117 12652 1k5 5% 2W  
3513▲ 4822 051 20008 0Ω JUMPER  
3517 4822 117 11846 10k 5% 1/16W  
3518▲ 4822 116 83027 R22 5% 3W

3518 4822 117 10422 0.33Ω 5% 3W  
3520 4822 117 11149 82k 1% 0.1W  
3521 4822 116 52219 330Ω 5% 0.5W  
3525▲ 4822 052 10229 22Ω 5% 0.33W  
3528 4822 116 83868 150Ω 5% 0.5W  
3529 4822 117 11778 4Ω 7% 5%  
3530 4822 050 13902 3k9 1% 0.4W  
3532▲ 4822 051 20008 0Ω JUMPER  
3534 4822 051 20224 220k 5% 0.1W  
3536 4822 051 20393 39k 5% 0.1W

3537 4822 117 11846 10k 5% 1/16W  
3538 4822 050 11004 100k 1% 0.4W  
3539 4822 116 52251 18k 5% 0.5W  
3540 4822 101 11189 4.7k 30% 0.1W  
3541 4822 117 12653 47Ω 5% 2W  
3542▲ 4822 053 21475 4M7 5% 0.5W  
3545▲ 4822 053 21225 2M2 5% 0.5W  
3546▲ 4822 053 21475 4M7 5% 0.5W  
3565 4822 117 11846 10k 5% 1/16W  
3566 4822 051 20331 330Ω 5% 0.1W

3567 4822 051 20681 680Ω 5% 0.1W  
3568 4822 051 20101 100Ω 5% 0.1W  
3569▲ 4822 051 20102 1k 5% 0.1W  
3601 4822 116 90885 8k2 X 6  
3602 4822 117 12168 2k2 X 6  
3603 4822 116 90884 8k2 X 10  
3610 4822 117 11846 10k 5% 1/16W  
3612 4822 051 20224 220k 5% 0.1W  
3613▲ 4822 051 20008 0Ω JUMPER  
3614▲ 4822 051 20109 10Ω 5% 0.1W

3614▲ 4822 051 20153 15k 5% 0.1W  
3615▲ 4822 051 20109 10Ω 5% 0.1W  
3615 4822 117 11846 10k 5% 1/16W  
3616▲ 4822 051 20109 10Ω 5% 0.1W  
3616 4822 051 20223 22k 5% 0.1W  
3617 4822 050 11203 12k 1% 0.4W  
3618 4822 050 11503 15k 1% 0.4W  
3620 4822 050 11001 100Ω 1% 0.4W  
3621 4822 051 20561 560Ω 5% 0.1W  
3622 4822 051 20561 560Ω 5% 0.1W

3623 4822 117 11846 10k 5% 1/16W  
3624 4822 051 20101 100Ω 5% 0.1W  
3625 4822 051 20101 100Ω 5% 0.1W  
3626 4822 050 11001 100Ω 1% 0.4W  
3627 4822 050 11001 100Ω 1% 0.4W  
3628 4822 051 20101 100Ω 5% 0.1W  
3630 4822 051 20822 8k2 5% 0.1W  
3630 4822 117 11383 12k 1% 0.1W  
3631 4822 117 10834 47k 1% 0.1W  
3632 4822 051 20333 33k 5% 0.1W

3633 4822 051 20333 33k 5% 0.1W  
3634 4822 117 10834 47k 1% 0.1W  
3635 4822 051 20154 150k 5% 0.1W  
3636 4822 117 10834 47k 1% 0.1W  
3638 4822 050 11202 1k2 1% 0.4W  
3638 4822 116 52249 1k8 5% 0.5W  
3639 4822 051 20562 5k6 5% 0.1W  
3640 4822 050 18202 8k2 1% 0.4W  
3641 4822 050 18202 8k2 1% 0.4W  
3642 4822 050 11001 100Ω 1% 0.4W

3650 4822 117 11449 2k2 1% 0.1W  
3653 4822 051 20105 1M 5% 0.1W  
3654 4822 051 20822 8k2 5% 0.1W  
3655 4822 050 11001 100Ω 1% 0.4W  
3656 4822 117 11503 220Ω 1% 0.1W  
3657 4822 050 11001 100Ω 1% 0.4W  
3658 4822 051 20681 680Ω 5% 0.1W  
3660 4822 050 11001 100Ω 1% 0.4W  
3661 4822 050 11001 100Ω 1% 0.4W  
3662 4822 050 11002 1k 1% 0.4W

3663 4822 051 20681 680Ω 5% 0.1W  
3664 4822 051 20104 100k 5% 0.1W  
3666 4822 051 20273 27k 5% 0.1W  
3668 4822 051 20224 220k 5% 0.1W  
3669 4822 051 20101 100Ω 5% 0.1W  
3670 4822 050 11001 100Ω 1% 0.4W  
3674 4822 116 52283 4k7 5% 0.5W  
3676 4822 050 12703 27k 1% 0.4W  
3677 4822 050 18202 8k2 1% 0.4W  
3681 4822 117 11846 10k 5% 1/16W

3682 4822 117 11846 10k 5% 1/16W  
3683 4822 050 11004 100k 1% 0.4W  
3684 4822 117 11846 10k 5% 1/16W  
3685 4822 116 83884 47k 5% 0.5W  
3686▲ 4822 051 20153 15k 5% 0.1W  
3690 4822 051 20182 1k8 5% 0.1W  
3690 4822 117 11454 820Ω 1% 0.1W  
3694 4822 051 20562 5k6 5% 0.1W  
3695 4822 051 20562 5k6 5% 0.1W  
3696 4822 051 20562 5k6 5% 0.1W

3697 4822 116 52213 180Ω 5% 0.5W  
3698▲ 4822 051 20102 1k 5% 0.1W  
3996▲ 4822 051 20153 15k 5% 0.1W  
3996 4822 117 11846 10k 5% 1/16W  
3997 4822 051 20154 150k 5% 0.1W  
3997 4822 051 20223 22k 5% 0.1W  
3997 4822 051 20562 5k6 5% 0.1W  
3997 4822 051 20683 68k 5% 0.1W  
3997 4822 117 10834 47k 1% 0.1W

5014 4822 157 63065 0.68μH 10%  
5015 4822 152 20547 0.68μH 10%  
5100▲ 4822 157 53941 100μH 10%  
5206 4822 157 53303 12μH 10%  
5206 4822 157 53634 5.6μH 10%  
5209 4822 157 52333 100μH 10%  
5260 4822 157 70704 38.9 MHz  
5260 4822 157 70942 45.75MHz  
5286 4822 157 53303 12μH 10%  
5287 4822 157 53303 12μH 10%

5288 4822 157 53303 12μH 10%  
5442 4822 157 53139 4.7μH 10%  
5445▲ 4822 140 10612 L.O.T.  
5451 4822 158 10549 12μH 10%  
5457 4822 157 11167 47μH 5%  
5458 4822 157 11167 47μH 5%  
5500 4822 157 10999 LINE FILT.30mH  
5500 4822 157 11163 LINE 22mH  
5515 4822 157 60171 EMI FILT.100MHz  
5516 4822 157 60171 EMI FILT.100MHz

5540 4822 157 52007 4U7 10%  
5545▲ 4822 146 10716 S.M.TRAFO  
5550 4822 157 60171 EMI FILT.100MHz  
5551 4822 157 71157 27μH 5%  
5570 4822 157 60171 EMI FILT.100MHz  
5571▲ 4822 157 51462 10μH 10%  
5573 4822 157 60171 EMI FILT.100MHz  
5601 4822 157 60123 6.8μH 10%  
5602 4822 157 60123 6.8μH 10%  
5620 4822 157 60123 6.8μH 10%

5680 4822 157 52983 22μH 10%  
5683 4822 157 60123 6.8μH 10%  
5690 4822 157 60123 6.8μH 10%



6102▲ 4822 130 30621 1N4148  
6110▲ 4822 130 30621 1N4148  
6111▲ 4822 130 30621 1N4148  
6144 4822 130 34382 BZX79-C8V2  
6151 4822 130 34382 BZX79-C8V2  
6251▲ 4822 130 30621 1N4148  
6254 4822 130 34233 BZX79-C5V1  
6402▲ 4822 130 30621 1N4148  
6403 4822 130 42488 BYD33D  
6424 4822 130 34382 BZX79-C8V2

6440 4822 130 42488 BYD33D  
6441 4822 130 42488 BYD33D  
6443 4822 130 42488 BYD33D  
6444 4822 130 34145 BZX79-B39C  
6445▲ 4822 130 32896 BYD33M  
6449▲ 4822 130 42489 BYD33G  
6454 5322 130 31938 BYV27-200  
6461▲ 4822 130 30621 1N4148  
6464▲ 4822 130 30621 1N4148  
6470▲ 4822 130 42489 BYD33G

6480 4822 130 34382 BZX79-C8V2F  
6500 4822 130 34328 BZX79-B30  
6501 4822 130 34328 BZX79-B30  
6502 4822 130 31083 GP15K-16  
6502▲ 4822 130 80858 1N5062  
6503 4822 130 31083 GP15K-16  
6503▲ 4822 130 80858 1N5062  
6504 4822 130 31083 GP15K-16  
6504▲ 4822 130 80858 1N5062  
6505 4822 130 31083 GP15K-16

6505▲ 4822 130 80858 1N5062  
6507 4822 130 42606 BYD33J  
6508 4822 130 42606 BYD33J  
6524▲ 4822 130 31631 BYV10-20  
6537 4822 130 30842 BAV21  
6540 4822 130 30842 BAV21  
6550 4822 130 10807 BYM36C

6560 5322 130 31938 BYV27-200  
6563 4822 130 34233 BZX79-C5V1  
6570 5322 130 31938 BYV27-200

6600▲ 4822 130 34173 BZX79-C5V6  
6610 4822 130 34142 BZX79-B33  
6610 4822 130 82037 HZT33  
6650 4822 130 30862 BZX79-C9V1  
6653 4822 130 34233 BZX79-C5V1  
6663▲ 4822 130 82029 LTL307P



7001 5322 130 41983 BC858B  
7002 5322 130 41983 BC858B  
7003 5322 130 41983 BC858B  
7004 5322 130 41983 BC858B  
7102▲ 5322 130 41982 BC848B  
7109▲ 5322 130 41982 BC848B  
7110▲ 5322 130 41982 BC848B  
7116▲ 5322 130 41982 BC848B  
7120 4822 209 90462 TDA7056B  
7124▲ 5322 130 41982 BC848B

7150▲ 5322 130 41982 BC848B  
7183▲ 5322 130 41982 BC848B  
7214 5322 130 41983 BC858B  
7215▲ 5322 130 41982 BC848B  
7216▲ 5322 130 41982 BC848B  
7217▲ 5322 130 41982 BC848B  
7225 4822 209 15105 TDA8363  
7225 4822 209 15106 TDA8361E  
7225 4822 209 15251 TDA8362E  
7225 4822 209 15285 TDA8360E

7240▲ 5322 130 41982 BC848B  
7245 4822 209 90129 TDA8395P  
7255 4822 209 12635 TDA4665  
7269▲ 5322 130 41982 BC848B  
7310 4822 130 41782 BF422  
7330 4822 130 41782 BF422  
7350 4822 130 41782 BF422  
7401 4822 209 60955 TDA3653B  
7440 4822 130 60511 BC847B  
7441 5322 130 44647 BC368

7445 4822 130 10206 BUT11AX  
7480 4822 130 40855 BC337  
7518 4822 130 10806 STP6NA60FI  
7518 4822 130 63787 STP4NA60FI  
7520▲ 4822 209 90025 MC44603P  
7565 4822 130 40937 BC548B  
7566 5322 130 41983 BC858B  
7600 4822 209 14646 SAA5290ZP  
7608▲ 5322 130 41982 BC848B  
7610▲ 4822 209 73852 PMBT2369

7620 4822 209 90962 ST24W04B1  
7650▲ 5322 130 41982 BC848B  
7667▲ 5322 130 41982 BC848B  
7677 4822 130 42705 BC847  
7681▲ 5322 130 41982 BC848B  
7682▲ 5322 130 41982 BC848B

## Picture tube panel [B1]

## Various

▲ 4822 255 10355 CRT SOCKET  
▲ 4822 255 70293 CRT SOCKET 14"  
1010 4822 212 11132 CRT PANEL 14"  
1010 4822 212 11133 CRT PANEL 20"



2304 4822 126 13461 680pF 10% 50V  
2344 4822 126 13461 680pF 10% 50V



3300 4822 117 11683 2k2 LIN POTM  
3303 4822 116 52219 330Ω 5% 0.5W  
3303 4822 116 83883 470Ω 5% 0.5W  
3304 4822 051 20129 12Ω 5% 0.1W  
3304 4822 051 20189 18Ω 5% 0.1W  
3311 4822 116 83883 470Ω 5% 0.5W  
3311 4822 050 18201 820Ω 1% 0.4W  
3312 4822 050 11001 100Ω 1% 0.4W  
3313 4822 050 11001 100Ω 1% 0.4W  
3321▲ 4822 116 83749 680Ω 1/4W

3322 4822 050 11502 1k5 1% 0.4W  
3323 4822 116 52219 330Ω 5% 0.5W  
3323 4822 116 83883 470Ω 5% 0.5W  
3324 4822 051 20569 56Ω 5% 0.1W  
3324 4822 051 20339 33Ω 5% 0.1W  
3331 4822 116 83883 470Ω 5% 0.5W  
3331 4822 050 18201 820Ω 1% 0.4W  
3332 4822 050 11001 100Ω 1% 0.4W  
3333 4822 050 11001 100Ω 1% 0.4W  
3340 4822 117 11683 2k2 LIN POTM

3343 4822 116 52219 330Ω 5% 0.5W  
3343 4822 116 83883 470Ω 5% 0.5W  
3344▲ 4822 051 20109 10Ω 5% 0.1W  
3344 4822 051 10159 15Ω 2% 0.25W  
3351 4822 116 83883 470Ω 5% 0.5W  
3351 4822 050 18201 820Ω 1% 0.4W  
3352 4822 050 11001 100Ω 1% 0.4W  
3353 4822 050 11001 100Ω 1% 0.4W  
3357 4822 117 11449 2k2 1% 0.1W  
3357 4822 051 20122 1k2 5% 0.1W

3360 4822 051 20561 560Ω 5% 0.1W  
3361 4822 050 13302 3k3 1% 0.4W  
3362 4822 051 20681 680Ω 5% 0.1W



5370 4822 157 51157 3.3μH 10%  
5370 4822 157 50961 22μH 10%



6310 4822 130 34174 BZX79-C4V7  
6314 4822 130 30842 BAV21  
6330 4822 130 34174 BZX79-C4V7  
6334 4822 130 30842 BAV21  
6350 4822 130 34174 BZX79-C4V7  
6354 4822 130 30842 BAV21



7300 4822 130 40938 BC548  
7320 4822 130 40938 BC548  
7340 4822 130 40938 BC548  
7360 4822 130 40941 BC558